



## ***FASTCAMERA SERIES***

**FASTCAMERA40  
USER'S MANUAL**

**FVM-00400**

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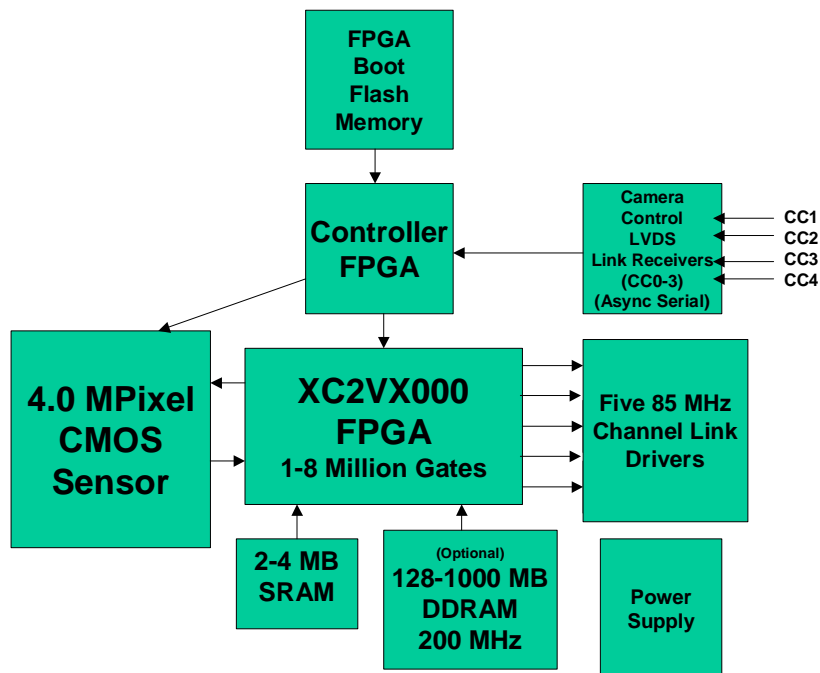
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## 1 INTRODUCTION

The FastCamera 40 is a 4 megapixel CMOS camera with internal memory and FPGA's that enable it to do real-time processing. Thus it is what one would term a "smart" camera. The standard programming that is supplied with the base camera forms the basis of the most used and demanded function for data processing. It is expected that each customer may wish to customize this programming for their use. This can be achieved by purchasing additional IP from FastVision or in-house development. Inquires or discussions are always welcome.

## 2 FEATURES AND SPECIFICATIONS FASTCAMERA 40

The block diagram below shows the major subsystems in the camera. The camera is designed to support many different applications by customization of the programmable logic in the camera. FastVision or the customer can customize the size and content of the FPGAs and memory in the camera to contain many different features, processing algorithms, and storage schemes. This manual discusses the 'Standard' version of the camera. Customization of the FPGAs in the camera, requires significant support from FastVision. Please contact FastVision for a quote for the development tools and support.



- The FastCamera40 uses a 2352H x 1728V (4 megapixel) CMOS digital image sensor capable of 240 frames/second operation at full resolution
- 2352H x 1728V image resolution
- 7-micron-square active-pixel photodiodes
- 240+ frames per second, progressive-scan

- Maximum Frame Rate 400,000 fps at 1 x 2048 pixels
- Monochrome or Bayer pattern color
- Up to Five Channel links in three MDR26 connectors for the full frame rate
- Camera link Basic, Medium, Full and Full+(4, and 5 CL) supported
- FPGA and memory-based configurable interface formats and onboard processing
- Supported by a full range of software tools
- Digital binning in order to achieve increased Signal to Noise ratios at full frame rates
- Optional SRAM for ultra fast processing
- Optional additional DDRAM and increased FPGA size for additional processing capability
- Trigger-able electronic rolling shutter
- C mount lens

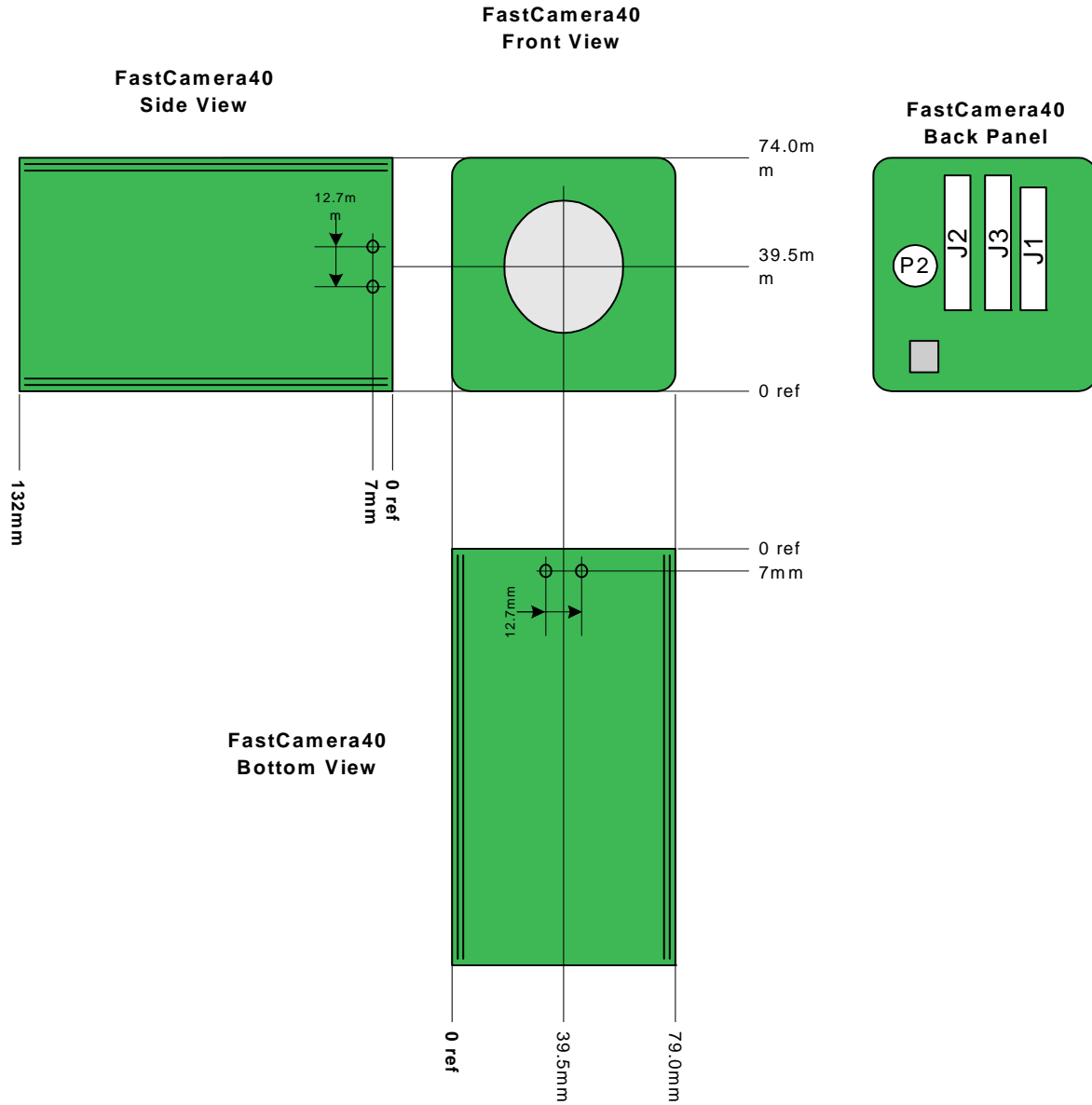
## **2.1 FastCamer40 Image Sensor Specification**

- Uses Micron Imaging's MI-MV40 sensor
- 2352 x 1728 x 8 bits @ 240 fps (10 bits 240 fps)
- 15.36 mm x 12.29 mm active area
- 7-micron square active pixels
- 40% Fill Factor
- Monochrome or color (Bayer Pattern)
- On-chip Noise Cancellation
- Dynamic range 59 db
- Monochrome: 2500 bits per lux-second @ 550 nm
- Shutter 99.9% efficiency
- Noise 58 db (10 bit mode lowest sensor gain setting, nominal pixel of 512 counts)
- Sixteen (16) parallel output ports
- On-chip 10-bit analog-to-digital converters



## 2.2 Physical Specifications

### FASTCAMERA40 CASE AND MOUNTING DIMENSIONS





## 2.3 Connectors

### 2.3.1 Power Connector HR10A-10R-12PB

Pin	Function
1	Ground
2	+5 Volts
3	Ground
4	Reserved for application dependent I/O
5	Ground
6	+5 Volts
7	Reserved for application dependent I/O
8	Ground
9	+5 Volts
10	Reserved for application dependent I/O
11	+5 Volts
12	Reserved for application dependent I/O

### 2.3.2 Data Connector J1

Pin	Signal	Pin	Signal
1	Ground	14	Ground
2	CL1_TXOUT0N	15	CL1_TXOUT0P
3	CL1_TXOUT1N	16	CL1_TXOUT1P
4	CL1_TXOUT2N	17	CL1_TXOUT2P
5	CL1_TXCLKOUTN	18	CL1_TXCLKOUTP
6	CL1_TXOUT3N	19	CL1_TXOUT3P
7	SERTCP	20	SERTCN
8	CC1N	21	SERTFGP
9	CC2P	22	CC1P
10	CC3N	23	CC2N
11	CC4P	24	CC3P
12	SERTFGN	25	CC4N
13	Ground	26	Ground

### 2.3.3 Data Connector J2

Pin	Signal	Pin	Signal
1	Ground	14	Ground
2	CL2_TXOUT0N	15	CL2_TXOUT0P
3	CL2_TXOUT1N	16	CL2_TXOUT1P
4	CL2_TXOUT2N	17	CL2_TXOUT2P
5	CL2_TXCLKOUTN	18	CL2_TXCLKOUTP
6	CL2_TXOUT3N	19	CL2_TXOUT3P
7	Application Specific I/O (P)	20	Application Specific I/O (N)
8	CL3_TXOUT0N	21	CL3_TXOUT0P
9	CL3_TXOUT1N	22	CL3_TXOUT1P
10	CL3_TXOUT2N	23	CL3_TXOUT2P
11	CL3_TXCLKOUTN	24	CL3_TXCLKOUTP
12	CL3_TXOUT3N	25	CL3_TXOUT3P

13	Ground	26	Ground
----	--------	----	--------

### 2.3.4 Data Connector J3

Pin	Signal	Pin	Signal
1	Ground	14	Ground
2	CL4_TXOUT0N	15	CL4_TXOUT0P
3	CL4_TXOUT1N	16	CL4_TXOUT1P
4	CL4_TXOUT2N	17	CL4_TXOUT2P
5	CL4_TXCLKOUTN	18	CL4_TXCLKOUTP
6	CL4_TXOUT3N	19	CL4_TXOUT3P
7	Application Specific I/O (P)	20	Application Specific I/O (N)
8	CL5_TXOUT0N	21	CL5_TXOUT0P
9	CL5_TXOUT1N	22	CL5_TXOUT1P
10	CL5_TXOUT2N	23	CL5_TXOUT2P
11	CL5_TXCLKOUTN	24	CL5_TXCLKOUTP
12	CL5_TXOUT3N	25	CL5_TXOUT3P
13	Ground	26	Ground

## 3 **POWER REQUIREMENTS**

Power requirements are a strong function of the application, the camera is 15 Watts worst case, 5 to 10 Watts typical. Low noise + 5 Volt input recommended. Internally the camera has high frequency switching supplies that convert the 5 volt input to 3.3, 2.5 and 1.8 volts.

## 4 **TIMING**

- Camera Clock is configurable from 25 to 85 MHz depending on the application.
- Pixel data is valid when LVAL, FVAL and DVAL are true.
- Minimum two clocks LVAL and FVAL inactive between lines and frames.
- CC4 reserved for application specific needs.

## 5 **TRIGGER MODES**

Trigger modes are selected by serial commands.

- Free Running, camera generates frames without triggers.
- CC1 Positive edge triggered.
- CC1 When active, expose, falling edge reads out the sensor.

## 6 **READ OUT MODES**

The camera can be read out with 1,2,3,4, or 5 camera links, via the serial port (for the really patient), or via the USB port. Only image data from the selected ROI is sent. The camera can be configured to read out when the image is taken, to read out from memory on serial command or the CC2 positive edge.

**Note:** The modes that use 3 channel links or less are according to the Camera Link Standard version 1.0.

**Data formats supported:**

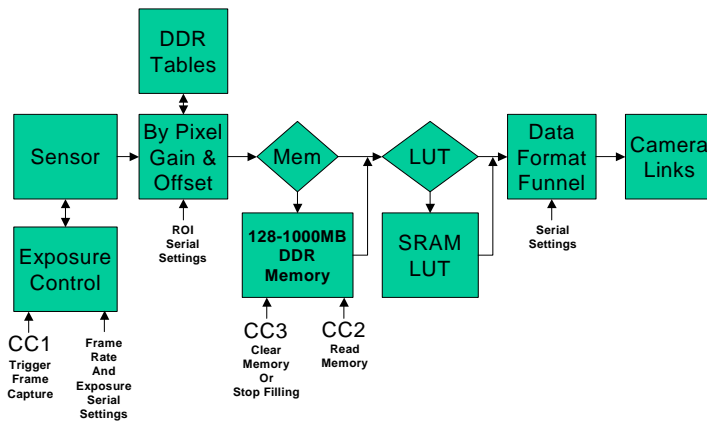
Mode	Camera Link Format MSB ----- LSB	Read Out Mode
0	CL1_A[7:0]	Single tap 8 bits (Basic) (default)
1	CL1_B[1:0],CL1_A[7:0]	Single tap 10 bits (Basic)
2	CL1_A[7:0] even pixels (0,2,...) CL1_B[7:0] odd pixels (1,3,...)	Two Taps 8 bits (Basic)
3	CL1_B[1:0],CL1_A[7:0] even CL1_B[5:4],CL1_C[7:0] odd	Two Taps 10 bits (Basic)
4	CL1_A[7:0] Red, Y CL1_B[7:0] Green, U CL1_C[7:0] Blue, V	Three Taps 8 bits (Color) (Basic)
5	CL1_B[1:0],CL1_A[7:0] Red, Y CL1_B[5:4],CL1_C[7:0] Green, U CL2_D[1:0],CL2_C[7:0] Blue, V	Three Taps 10 bits (Color) (Medium)
6	CL1_B[7:0],CL1_A[7:0] CL1_A[4:0] Red, U CL1_B[2:0],CL1_A[7:5] Green,Y CL1_B[7:3] Blue V	One tap 16 Bits RGB565
7	CL1_A[7:0] pixels $4*n+0$ CL1_B[7:0] pixels $4*n+1$ CL1_C[7:0] pixels $4*n+2$ CL2_A[7:0] pixels $4*n+3$	Four Taps 8 bits (Medium)
8	CL1_B[1:0],CL1_A[7:0] $4*n+0$ CL1_B[5:4],CL1_C[7:0] $4*n+1$ CL2_C[1:0],CL2_B[7:0] $4*n+2$ CL2_C[4:5],CL2_A[7:0] $4*n+3$	Four Taps 10 bits (Medium)
9	CL1_A[7:0] pixels $5*n+0$ CL1_B[7:0] pixels $5*n+1$ CL1_C[7:0] pixels $5*n+2$ CL2_A[7:0] pixels $5*n+3$ CL2_B[7:0] pixels $5*n+4$	Five Taps 8 bits (Medium)
10	CL1_B[1:0],CL1_A[7:0] $5*n+0$ CL1_B[5:4],CL1_C[7:0] $5*n+1$ CL2_C[1:0],CL2_B[7:0] $5*n+2$ CL2_C[4:5],CL2_A[7:0] $5*n+3$ CL3_C[1:0],CL3_B[7:0] $5*n+4$	Five Taps 10 bits (Full)
11	CL1_A[7:0] R even CL1_B[7:0] G even CL1_C[7:0] B even CL2_A[7:0] R odd CL2_B[7:0] G odd CL2_C[7:0] B odd	Six Taps 8 bits (Color) (Medium)
12	CL1_B[1:0],CL1_A[7:0] R even CL1_B[5:4],CL1_C[7:0] G even CL2_C[1:0],CL2_B[7:0] B even CL2_C[4:5],CL2_A[7:0] R odd CL3_C[1:0],CL3_B[7:0] G odd CL3_C[5:4],CL3_A[7:0] B odd	Six Taps 10 bits (Color) (Full)
13	CL1_A[7:0] $8*n+0$ CL1_B[7:0] $8*n+1$ CL1_C[7:0] $8*n+2$ CL2_A[7:0] $8*n+3$ CL2_B[7:0] $8*n+4$ CL2_C[7:0] $8*n+5$ CL3_A[7:0] $8*n+6$	Eight Taps 8 bits (Full)

	CL3_B[7:0] 8*n+7	
14	CL1_B[1:0],CL1_A[7:0] 8*n+0 CL1_B[5:4],CL1_C[7:0] 8*n+1 CL2_B[1:0],CL2_A[7:0] 8*n+2 CL2_B[5:4],CL2_C[7:0] 8*n+3 CL3_B[1:0],CL3_A[7:0] 8*n+4 CL3_B[5:4],CL3_C[7:0] 8*n+5 CL2_FVAL,CL2_LVAL,CL2_SP,CL1_SP, CL1_B[7:6],CL1_B[3:2] 8*n+6 CL3_FVAL,CL3_LVAL,CL1_DVAL,CL3_SP , CL2_B[7:6],CL2_B[3:2] 8*n+7	Eight Taps 10 bits (Full)
15	CL1_A[7:0] 16*n+0 CL1_B[7:0] 16*n+1 CL1_C[7:0] 16*n+2 CL2_A[7:0] 16*n+3 CL2_B[7:0] 16*n+4 CL2_C[7:0] 16*n+5 CL3_A[7:0] 16*n+6 CL3_B[7:0] 16*n+7 CL3_C[7:0] 16*n+8 CL4_A[7:0] 16*n+9 CL4_B[7:0] 16*n+10 CL4_C[7:0] 16*n+11 CL5_A[7:0] 16*n+12 CL5_B[7:0] 16*n+13 CL5_C[7:0] 16*n+14 CL2_FVAL,CL3_FVAL,CL4_FVAL,CL5_SP, CL4_SP,CL3_SP,CL2_SP,CL1_SP 16*n+15	16 Taps 8 bits (5 Camera Links)

16	CL1_B[1:0],CL1_A[7:0] R 4*n+0 CL1_B[5:4],CL1_C[7:0] G 4*n+0 CL2_B[1:0],CL2_A[7:0] B 4*n+0 CL2_B[5:4],CL2_C[7:0] R 4*n+1 CL3_B[1:0],CL3_A[7:0] G 4*n+1 CL3_B[5:4],CL3_C[7:0] B 4*n+1 CL4_B[1:0],CL3_A[7:0] R 4*n+2 CL4_B[5:4],CL3_C[7:0] G 4*n+2 CL5_B[1:0],CL3_A[7:0] B 4*n+2 CL5_B[5:4],CL3_C[7:0] R 4*n+3 CL1_SP,CL1_B[7:6],CL1_[3:2], CL2_SP,CL2_B[7:6],CL2_[3:2] G 4*n+3 CL3_SP,CL3_B[7:6],CL3_[3:2], CL4_SP,CL4_B[7:6],CL4_[3:2] B 4*n+3	12 Taps 10 Bits (Color) (5 Camera Links)
17	CL1_B[1:0],CL1_A[7:0] CL1_B[5:4],CL1_C[7:0] CL2_B[1:0],CL2_A[7:0] CL2_B[5:4],CL2_C[7:0] CL3_B[1:0],CL3_A[7:0] CL3_B[5:4],CL3_C[7:0] CL4_B[1:0],CL3_A[7:0] CL4_B[5:4],CL3_C[7:0] CL5_B[1:0],CL3_A[7:0] CL5_B[5:4],CL3_C[7:0]	10 Taps 10 Bits FastCamera13
254	NA	Use Serial Port
255	NA	Use USB

Other modes are possible with custom FPGA configurations purchased from FastVision or developed in-house.

## 7 STANDARD CAMERA DATA FLOW



DDR Memory can be organized as a FIFO or as a circular buffer. CC3 is used to clear the memory (FIFO) or stop filling the memory (circular buffer). CC2 is used to read the top most image from memory (FIFO) or to read the oldest non-read image from memory (Circular buffer).

The LUT converts the input pixels (10 Bits) to the output pixel size (8 or 10 Bits). These tables can also be used to improve the linearity of the sensor, or to set the gamma of the camera.

DDR Memory can be used for image averaging. In this mode images are summed into a 32 bit per pixel buffer, which can be read out at any time (CC2) and/or cleared (CC3). CC1, CC2, and CC3 can be replaced by Serial Commands, or can be always enabled or disabled for free running operation.

## 8 THE STANDARD CAMERA FUNCTIONALITY

The 'Standard' Camera is a set of FPGAs designed to support most of the typical uses of the camera. There is functionality that you will find is a good starting point for modifications. This design is copyrighted IP from FastVision and forms the reference design, which is available from FastVision for use only with the FastCamera40.

The frame rate and exposure can be controlled in several ways:

### 8.1 Exposure Control

- Free Running Mode
- CC1 positive(or negative) edge triggered
- CC1 exposure control (positive or negative level triggered)

In Free Running Mode, the frame rate and exposure time is set via the serial port and the camera acquires images without further outside control.

In CC1 edge triggered mode, the camera is triggered by the selected edge polarity to expose for the time selected by the serial setting. Each edge of the CC1 input will cause the acquisition of an image. If CC1 is toggled faster than the exposure and readout rate allows, the additional triggers are ignored. Frame readout may overlap exposure in both sensors. The FastCamera40 has a rolling shutter. The shutter speed granularity is approximately 2 microseconds. Finer resolutions are possible see below.

In CC1 level triggered mode, the exposure is set by the active time of the CC1 input. When CC1 is asserted (programmable level high or low) the sensor is reset and integration begins. When CC1 is de-asserted the sensor is read out. The exposure ends at the end of the line in which CC1 falls, or the end of the next line.

### **8.1.1 Rolling shutter**

The FastCamera40 supports rolling shutter mode. The rolling shutter allows the camera to operate faster, by resetting each line a programmable delay (the exposure time) before it is read out. These two CMOS sensors use photo diodes as the active elements of the sensor. When an exposure is started the diodes are emptied out by a reset process, which clears all the stored charge from each diode. This reset process starts the exposure for a particular pixel. After the exposure is complete, the charge in the photo diode is read out to an amplifier and converter section in the sensor. This readout process also resets the charge in the diode.

All the pixels in a line are readout and reset, or just reset at the same time. The rolling reset process takes advantage of the readout process to set the exposure. When the sensor is read out each line in turn is dumped into the converter to read out all the pixels in a line. The sensor contains separate amplifiers and converters for each column in the sensor (for the FastCamera 40 that is 2352 converters!). At the same time a line is read out and reset from the sensor a different line can be reset. By selecting the line to be reset, to be a line to be read out and reset in the near future, the exposure time is set to the number of line times between the current line, and the future line being reset. While this process allows exposure to occur while frame data is being read out, it is at a price. If your light source is continuous, you will see a compression or extension of the moving object that moves in the opposite or same direction as the readout. Typically this compression is only apparent at very high frame speeds, and / or object speeds. If your light source is a flash the compression will not occur as you can make sure all the lines are exposing, before the flash.

The fastest a line can be read out from the sensors is in the time it takes for the converters to convert the charge from the photodiodes, into digital values. In this sensor this is 132 clocks. In the standard camera the sensor clock rate is 66.66 MHz, this means the fastest the lines can be sampled is 1.98 microseconds per line. During this time a partial line of 2048 pixels can be read out. In the FastCamera40 the additional pixels extend the line time to 153 clocks or 2.29 microseconds per line. (Strictly speaking the converter time is 128 clocks, but there is additional overhead in reading out the whole line which brings the time to 153 clocks, please see the data sheets of the sensors for the details.) If the number of columns in the ROI is less than 2048 in the FastCamera40 this time is reduced to 132 clocks per line or 1.98 microseconds. The sensor can not be read out faster than 1.98 microseconds per line.



### 8.1.2 Shutter Speeds

The granularity of the shutter speed at the highest readout rate possible is 1.98 microseconds. This is not the whole story though. You may lengthen the line time by programming the clocks per line register (see serial command 0x04 below). By slowing down the line rate, you can dial the shutter speed (at the cost of maximum frame rate), to any time above 132 clocks. So to get a shutter speed of 3.0 microseconds you would set the clocks per line to 200 clocks. The highest full frame readout rate would be 192 full frames per second.

## 8.2 Pixel Gain and Offset

The offset and gain of each pixel can be calibrated by setting the pixel gain and offset arrays in the DDR memory of the camera. It is a good idea to do this as the large number of converters, need to be calibrated for best results.

**Note:** Only the pixels in the selected ROI are processed (see serial commands 0x02 and 0x03).

The offset table is set by taking several images with no light input, averaging them and uploading them to the offset (or dark field) table. These values are subtracted from the sensor pixel values (clipped so less than zero values are zero), before applying the gain table values.

The gain table values are obtained by taking a defocused image of a uniform object, with enough light to get to 75% of saturation at most on the brightest point in the image. Using the resulting image, gain values are computed (in the simplest case the pixel values and the ratio with 75% of full scale (191 (8 bit) or 768 (10 bits)) determines the gain.).

**Note:** It is assumed that you are trying for a uniformly lit field of view (i.e. a Flat Field). It is important that you have as nearly as is practical uniform field of light, as large or small gain values can introduce significant artifacts in your images.

The gain table contains 8 bit values, which are formatted unsigned 1.7 format, that is one binary digit followed by 7 fractional digits. This makes the range of gains 1.992 down to 0.00390625.

The equation for the pixel gain and offset is:

$$CP[i,j] = \text{Saturate}((2 * \text{Gain}[i,j] / 256) * \text{Clip}(P[i,j] - \text{Offset}[i,j]))$$

CP = Corrected Pixel

P = Raw sensor Pixel

Clip = 0 if  $P[i,j] - \text{Offset}[i,j]$  is negative, otherwise its  $P[i,j] - \text{Offset}[i,j]$ .

Saturate = FullScale if its input is greater than or equal to FullScale, otherwise it is its input value.

FullScale = 255 for 8 bit pixels 1023 for 10 bit pixels.

The camera has 120 MB (or more) of memory which can be used to store images, and do averaging. Each mode of operation is explained in the following sections. If the memory option is enabled, then sensor data goes to memory instead of down stream.

## 8.3 Memory Modes (Mem)

### 8.3.1 FIFO memory mode

In this mode memory is used as a first in first out memory (FIFO). The memory fills with images, until it is full, and then stops filling. At any time the user may request an image from the FIFO with CC2, or via a serial command. This will make room for a new image, which will be filled as soon as one is collected. This mode is typically used by systems

that can accept images in bursts, but can not accept a continuous stream of images. For example your system might take 8 images very fast and then read them out at a slower rate (via a single camera link for example), your average frame rate is slow, but your peak rate is high. If the host activates CC2 (or sends the serial command), it can cause the next image to be read out when it comes in (pre-trigger the readout).

### **8.3.2 Circular buffer memory mode**

In this mode the memory is used as a circular buffer. Each time an image is presented to the memory it is stored over-writing the oldest image in the buffer. When CC2 goes active (or via a serial command), the filling operation stops, after a selected number of frames are added to the memory buffer (delayed trigger).

After the memory image collection stops, each time CC2 toggles (or by serial command) the oldest remaining image is sent to the host. After the first CC2 is sent the host may send the next CC2 right away it does not have to wait for the filling to finish. Each image will be sent to the host as it becomes available.

After the memory is full the camera will not accept any more images until it is read out or reset via CC3 (or serial command). If the host is draining images as fast as they are coming in the image capture will never stop. A reset is required to re-arm the trigger and delay the process.

### **8.3.3 Image summing memory mode**

In this mode the memory is divided up into buffers the size of the selected ROI, but with 32 bit values. Each time an image is added to memory, it is added to the current average buffer. When a programmed count of images is exceeded, the system advances to the next buffer. When all the buffers are full, the system stops until it is read out (CC2) or reset (CC3).

**Note:** as each buffer is read out (CC2) it is reset. If the host reads out the buffers faster than they are collected, image collection will not stop. For example you can program the camera to total 10 images in each buffer, before passing it to the host. The host triggers the readout via CC2 (or serial command). The CC2 trigger may be sent at any time, the camera will provide a total buffer when it becomes available.

## **8.4 Lookup Table Option (LUT)**

If the lookup table option is enabled, the image data from up stream is passed through a lookup table. This applies a point transformation to each pixel value, producing 10 bit results, from the 8 or 10 bit input. 32 Bit totals bi-pass the block. (Or better you should not enable this block if you are doing image totaling, as it will only operate on the lower 10 bits of the total.)

## **8.5 Data Format Funnel**

The data format funnel takes the image data and parcels it out the camera links (or USB) as selected by the Format Funnel mode. If 8 bit data is selected the upper 8 bits of each pixel is sent. (Note if you don't like that use the LUT to change this behavior). 32 Bit data total buffers are sent as 8 bit data in little endian.

## **8.6 Internal Camera Memory**

The camera contains at least 128 MB of internal memory which can be expanded to 1000 MB, which can be used to FIFO the input images, to allow burst exposure, and slow readout operation. Only the image data in the selected ROI is stored. A serial command or the positive edge of CC3 can be used to clear FIFO memory. In addition, the camera contains LUTs for conversion from 10 to 10 bit, or 8 bit data which can be uploaded via the serial port.

## **9     SENSOR CONTROL**

### **Line Timing**

Sensor line and frame rates are controlled by the Camera State settings. The minimum line period depends on the sensor type and the ROI width. When the Line Period setting exceeds the minimum period, extra clocks are inserted between lines. The Line Valid period only depends on the ROI width and sensor type. The number of columns read from the sensor is always a multiple of 10 for the MV13 and 16 for the MV40 sensor. Since the Data FPGA also has access to the ROI settings it is possible for the actual ROI width (as sent to the frame grabber) to start and end on any pixel, however this is not implemented. The Control FPGA ensures that the starting and ending pixels of the ROI are read from the sensor. For the MV13 this means that as many as 9 pixels before and 9 pixels after the ROI could be read from the sensor, depending on the starting and stopping pixel values modulo 10. For the MV40 sensor as many as 15 pre- and post-ROI pixels could be read from the sensor, depending on the starting and stopping pixel values modulo 16. The software GUI is responsible for determining the actual number of pixels per line based on the sensor and readout mode.

Camera-Link or USB readout rate can be the limiting factor for the line rate in non-memory modes. In this case it is the GUI's responsibility to maintain the Line Period setting large enough to prevent FIFO overrun in the Data FPGA.

### **Frame Timing**

In free-running mode the minimum frame period depends on the ROI height and the line period. When either the frame period or exposure time setting exceeds the minimum period, extra clocks are inserted between frames. A 32-bit frame time counter allows frame periods from the minimum up to 64 seconds in increments of 15 nS.

In triggered modes the frame period is also affected by the exposure delay. In multiple trigger mode the exposure delay only affects the time until the first readout of the trigger sequence and subsequent frames run at the free-running rate as specified. In single-trigger mode with external exposure control, applying the trigger faster than the maximum rate will cause the camera to skip triggers and run at a submultiple of the trigger frequency. In multiple trigger mode and internal timed trigger mode, applying the trigger faster than the maximum rate will result in the camera running at the specified free-run rate.

Trigger pipelining is provided so that a trigger event which occurs during exposure or readout (just during readout in external exposure mode) will generate another frame.

### **Exposure Modes**

The MV13 sensor is normally used in TrueSNAP shutter mode. The sensor has analog frame storage and a transfer gate that allows exposure to overlap readout while exposing all lines together. The MV40 sensor does not have analog storage and can only be exposed in a rolling shutter mode. Thus the individual lines have different exposure windows which may or may not overlap depending on the line rate and exposure time settings. This is the equivalent of using the MV13 sensor with the transfer gate on all the time.

When viewing high-speed motion, the MV40 works best if the lines are read out at the maximum possible rate while additional time for exposure or frame period is inserted between frames. This maximizes the overlap in the line exposure times. Unfortunately in the non-memory readout modes with reduced output rates just the opposite must be done. Long interline delays in these modes causes excessive temporal

distortion in the moving image. This can appear as stretching or shrinking of vertically moving objects or trapezoidal distortion of horizontally moving objects.

## Trigger Modes

There are four basic trigger modes:

- Free-running mode ignores trigger inputs and reads out the sensor continuously at a programmed frame rate with programmed exposure timing.
- Multi-frame edge-triggered mode activates a programmed number of exposures after a programmed delay at a programmed frame rate with programmed exposure timing.
- Single edge-triggered mode activates a programmed exposure after a programmed delay.
- External exposure mode exposes during the active trigger period.

Free-running mode allows continuous readout at rates ranging from about one frame per minute to the maximum capability of the sensor. Exposure can be programmed in increments of 15 nS up to the frame period. There are gaps in the allowable exposure times due to interaction between the readout circuitry and pixel reset in the MV13 sensor. In the MV40 sensor the exposure time is even more restricted due to the rolling shutter mode of operation. The Control FPGA will generate the exposure timing as close as possible to the requested value.

Multi-frame edge-triggered mode effectively starts the camera into free-running mode for a programmed number of frames after a delay of zero to about 64 seconds programmable in 15 nS increments. The active edge of the trigger activates the time delay and the first exposure starts after this delay. The camera can be re-triggered during active readout. If the re-trigger event occurs too soon to start the subsequent exposure after the programmed delay, the next exposure will start as soon as possible thereafter. This prevents the effect of running at submultiples of the trigger rate when the trigger rate exceeds the capability of the camera.

Single edge-triggered mode is the same as asynchronous multi-frame edge-triggered mode with the frame count set to 1.

External exposure mode starts exposure as soon as possible after the active-going edge of the trigger and stops exposure as soon as possible after the inactive-going edge of the trigger. For the MV13, this provides an accurate exposure timing based on the trigger pulse width. For the MV40, limitations of the rolling shutter limit the exposure timing resolution to the line readout rate unless the exposure time exceeds the readout time. Subsequent exposures can overlap readout of the current frame in this mode. This can place further restrictions on the exposure timing resolution in both sensor types.

The MV13 camera has "synchronous" and "asynchronous" modes of exposure. These only apply to free-running and multi-frame modes. Single-frame and external modes are always "asynchronous." When the synchronous trigger mode bit is set, the exposure timing is linked to the readout timing to avoid exposure period jitter that can occur when the readout overlaps the exposure. In multi-frame triggered mode, the synchronous mode also creates an additional delay of one readout time that must be added to the programmed delay to find the actual delay to the first exposure.

## Trigger Options

There are four trigger sources - Camera Link CC1, TTL trigger input, Serial, and USB (via I2C).

- The CC1 and TTL sources can be enabled or disabled. They can also be active high (positive-going edge for edge-triggering or high level for external exposure) or active low (negative-going edge for edge-triggering or low level for external exposure).
- The Serial source uses the "O" command. This cannot be disabled in the Control FPGA. It must be disabled by the GUI if desired.
- The USB source uses writes to subaddress TBD. In external exposure mode two writes are required. Writes to subaddress TBD with bit 0 high start the exposure, and writes to subaddress TBD with bit 0

low end the exposure. This cannot be disabled in the Control FPGA. It must be disabled by the GUI if desired.

## Trigger Outputs

There are two trigger outputs, both TTL level on the P2 power connector. One output is high whenever the sensor is exposing. The other is high during the programmed delay period in the two edge-triggered modes.

## Reference Voltages

All sensor reference voltages are programmable and generated by two LTC1660 octal D/A converters. At power-up the DAC's are loaded with default values hard-coded into the Control FPGA. After the Data FPGA has been loaded the DAC's are updated with the settings in the default camera state storage page of flash memory.

Comparators in the Control FPGA check for changes in the DAC settings and re-load the DAC's whenever the values are changed. This can happen as a result of the host command to set camera state, or the host command to restore state from flash.

## Calibration

Both the MV13 and MV40 sensors have automated ADC calibration to reduce column-wise fixed-pattern noise. This is initiated at sensor reset or by using the CAL\_START\_N input. The Control FPGA always uses the sensor reset to initiate calibration. In addition, the MV40 sensor allows direct access to the internal calibration values on a serial interface consisting of the DATA\_CLK, DATA, RE\_N, and WE\_N pins. This version of the Control FPGA does not use this interface.

Automatic calibration is initiated after power-up when the sensor is released from reset. Reset is released before the Data FPGA is loaded to allow the sensor to stabilize during the load process. Reset is re-asserted for two clock cycles at the end of the initialization sequence, about TBD milliseconds after the reference voltage DAC's are updated from flash. This causes the sensor to re-calibrate with the new reference voltage settings.

After power-on initialization, the sensor is only calibrated on demand by the host using the Reset and Calibrate Sensor command. It is not automatically calibrated after changing the reference voltages. Thus the camera control GUI is responsible for periodic calibration as required.

## Power-on Initialization

The Control FPGA has a small embedded micro that runs through an initialization sequence at power-on. This same micro also implements the host, USB, and Data FPGA communication protocols and deals with flash memory and DAC's.

1. Immediately after power-on, the voltage reference DAC's are programmed with factory default values. This allows the sensor to stabilize under conditions close to the actual operating environment.
2. Page zero is read from the flash memory. This page holds information necessary to locate actual power-on data in the flash.
3. The sensor is released from reset and allowed to self calibrate and stabilize.
4. The Data FPGA is loaded from flash. If the offset or length stored in the header are not valid, the embedded micro skips to step 10. This may take several seconds depending on the FPGA size.
5. Camera State is read from one of the 8 pages as selected by the pointer in the header. If the pointer is not valid (1 - 8) or the selected page is uninitialized, the embedded micro skips to step 10.
6. Reference DAC's are reprogrammed with the updated values from flash.
7. Camera State is forwarded to the Data FPGA.

8. If the Pointer to the Data FPGA initialization area is valid and the length is valid and non-zero, the Data FPGA initialization is read from flash and forwarded to the Data FPGA.
9. The sensor is briefly reset to start another auto-calibration cycle.
10. The embedded micro enters the command service loop.

## 10 FLASH MEMORY

Table 2 shows the layout of flash memory. The first page is used only for main header information. Following this are eight pages for storing up to 8 camera states. Fast-Vision should reserve one of these for factory defaults. The Control FPGA will not write protect the defaults page so the Camera Control GUI should prevent the user from overwriting the factory settings, or it should have the factory default settings available in the software to restore them without using the flash. Any of these states can be chosen as the power-up default.

Flash Page Layout		
Page Offset	Pages	Description
0	1	Flash Memory Header Page
1	1	Camera State Storage 1
2	1	Camera State Storage 2
3	1	Camera State Storage 3
4	1	Camera State Storage 4
5	1	Camera State Storage 5
6	1	Camera State Storage 6
7	1	Camera State Storage 7
8	1	Camera State Storage 8
9	1370	FPGA bitstream (sized for 2V1000 + 2V250)
1379	2716	Available for Data FPGA Initialization and User Data
4095	1	Reserved by Atmel

Table 2 - Flash Memory Layout

Table 3 shows the layout of the first page in flash memory. This contains enough descriptors for the Camera Control GUI to determine the camera type and its options. This data should only be programmed by Fast-Vision. The camera GUI software must enforce this, as there is no write protection built into the Control FPGA firmware. Multibyte values are little endian.

Page Zero Flash Header		
Byte Offset	Bytes	Description
0	1	Flash Memory Part Type: 0x16 or 0x32 for 16 or 32 Mb flash
1	1	Flash Memory Revision ID: 0x01 for this revision
2	2	Offset in bytes to Camera ID data from top of page 0: 0x79 0x01
4	4	Flash page offset to FPGA data: 0x09 0x00 0x00 0x00
8	4	FPGA bitstream length in bytes
12	4	Flash page offset of Data FPGA initialization data
16	4	Data FPGA initialization data length in bytes
20	1	Camera State to load at power on (1 to 8)
20	357	Reserved for additional header / ID info
377	11	Part Number, ASCII "800??-5????"
388	3	Part Revision, ASCII "010"
391	9	Serial Number, ASCII "XXX?????"

400	128	FPGA bitstream file header from mkin, Null terminated string
-----	-----	--

Table 3 - Flash Memory Header Page

## 11 **CAMERA STATE STORAGE**

Internal to the Control FPGA all state is saved in a Block RAM. Copies of the current state can be saved to the flash or uploaded to the host. The current state can also be retrieved from flash or changed by the host. Only the host has random access to the camera state and this only when in setting state. Reading back the camera state always sends the entire state to the host. Table 4 shows the layout of the camera state memory. Except for sensor reference voltages, multibyte values are little endian.

Byte Offset (decimal)	Bytes	Description
0	4	C3, 5A, F0, 69 for detecting uninitialized buffers
4	2	VIn2: 14 D9
6	2	Vref1: 14 D9
8	2	Vtest: 20 00
10	2	Vref2: 23 E0
12	2	Vbias1: 30 00
14	2	Vref3: 32 E8
16	2	Vbias2: 40 00
18	2	Vref4: 41 36
20	2	Vbias3: 50 00
22	2	VIn1: 54 D9
24	2	Vbias4: 60 00
26	2	Vlp: 64 D9
28	2	Vunused1: 70 00
30	2	Vclamp3: 70 00
32	2	Vunused2: 80 00
34	2	Vrstpix: 8D 17
36	2	ROI Start Pixel
38	2	ROI End Pixel
40	2	ROI Start Line
42	2	ROI End Line
44	2	Line Period in Pixel Clocks
46	4	Exposure Time in Pixel Clocks
50	4	Frame Period in Pixel Clocks
54	4	Exposure Delay in Pixel Clocks
58	2	Serial Link Bit Period in Pixel Clocks
60	1	Camera Link Readout Mode
61	1	Camera Link Clock Frequency
62	1	Binning
63	1	Memory Options
64	1	Sensor Resolution - 8/10 bit
65	2	Trigger Mode (2nd Byte is Data FPGA dependent)
67	1	Frame count for Multi-Trigger mode
68	1	CC Mode: CC2, CC3, CC4 enable and edge select
69	59	Reserved for Control FPGA / base system extension
128	384	Available for Data FPGA functionality extensions

Table 4 - Camera State Memory Layout

Sensor reference voltages are presented to the DAC's exactly as they are stored in the state. The order listed above is the recommended order, but other orders may work. The format for these is MS byte first with the most significant nibble indicating the command code to the DAC. Pairs of values are sent, one to each DAC chip with the first going to U25 and the second to U26. Voltages are updated in the order sent. Default values shown are for the FastCamera40. For more information see the LTC1660 data sheet.

Whenever the state memory is updated from host or flash, the actual internal registers that implement the camera state change, too. Some of these are located in the Control FPGA and some in the Data FPGA. The Control FPGA forwards state data to the Data FPGA whenever it is updated.

The state memory holds all of the state variables currently defined for camera operational modes as well as some additional storage that can be defined as required for more sophisticated Data FPGAs. The Data FPGAs can count on this storage to be refreshed from flash after initial FPGA load and whenever the user restores state from one of the saved sets in flash.

In addition to the camera state storage pages, some of the flash memory is available for Data FPGA storage requirements such as pixel defect maps. The amount of flash available for this depends on the size of the Data FPGA and the size of the flash device. A pointer in the flash header in page zero indicates the starting page of the Data FPGA initialization area. Its length in pages (which may be zero) is stored in the flash header as well. Data from this initialization area is read out and sent to the data FPGA after initial FPGA load. Although the data in these pages has no predefined layout, the first page must start with the sequence 3C, A5, 0F, 96. This prevents transmission of uninitialized flash pages and serves to identify the following data as Data FPGA Initialization Data rather than Camera State.

## **12 SERIAL CONTROL INTERFACE**

### **Encoding**

Commands and response use the 7-bit ASCII code zero-extended to 8 bits. This is sometimes referred to as 8 bits, no parity or 7-bits space parity. For characters that require escape codes as listed below, the 8th bit is significant in the decoding, so only the normal character encoding needs to be escaped. For example hex 0d requires the escape sequence but hex 8d does not.

The data within a command or response packet may contain 8-bit binary data with escaping as described below. The camera requires only one stop bit for framing and sends one stop bit when responding.

### **Baud Rate**

To comply with the Camera Link specification, the default baud rate is 9600. The serial baud rate can be changed with a command on the serial link or via the USB port. To allow recovery after inadvertent setting of a higher baud rate than the host can handle, the serial link will revert to 9600 baud after 5 framing errors without receiving a valid command. The factory default startup baud rate is 9600, but the actual power-on baud rate can be changed via the user start-up settings.

### **Command Protocol**

Commands all begin with a character in the range g-z or G-Z where the case is ignored. Subsequent bytes of the command are command-specific, but most commands use hex characters 0-9 and a-f or A-F for the subsequent bytes. For commands using the hex encoding, all arguments are full bytes. Thus an odd number of hex digits is considered a protocol error. For commands that use binary encoding, the following character sequences are required:



<b>Character (hex, ASCII)</b>	<b>Sequence</b>
0d, <cr>	\ <cr>
5c, \	\\

All commands end in a non-escaped carriage-return, hex 0d.

When commands take arguments in hex, data must consist of pairs of hex characters representing full bytes. Within each byte the most significant nibble is sent first, but for multi-byte values the least significant byte is normally sent first. Thus when a command required a 32-bit value the string "23568912" would produce the hex value 12895623. Commands using hex arguments will ignore space characters, so "23568912" is equivalent to "23 56 89 12". Other characters will cause the command to abort and respond with negative acknowledgement at the next non-escaped carriage-return.

The intent of the protocol is to allow hand typing of commands from a terminal except for long commands like flash page write that would only be done using the Fast-Vision camera control GUI.

### **Response Protocol**

Each command signals completion by sending the command character G-Z (always uppercased) followed by any command-specific response and terminated with a non-escaped carriage-return, hex 0d. For commands that respond with binary data, the carriage-return (0d) and backslash (5c) are escaped as in the command protocol.

If a command cannot be completed for any reason the response will be the character ? (3F) and possibly a numeric error code, followed by a carriage-return to indicate negative acknowledgement.

Here again the intent is to use ASCII only response except for long data sequences like flash page read.

### **Synchronization**

The ASCII format of most commands combined with the escaped carriage-return in binary commands allows simple resynchronization after errors. Sending two carriage-returns will always reset the serial link to its default state, waiting for new command.

In addition, when the serial logic detects a framing error or protocol error, it will go into an idle state until the next non-escaped carriage-return. During the idle state any serial data is ignored. Any carriage-return which follows the idle state causes the negative acknowledgement sequence to be sent.

### **Timeout**

If the serial logic is in a state waiting for a command to be completed and no input is detected for 5 seconds, it will reset to the waiting for new command state without sending any response. This would typically happen after the serial link had noise from attaching or restarting the framegrabber.

\*\*\* TIMEOUT NOT IMPLEMENTED IN THIS VERSION \*\*\*

### **Command Forwarding**

All commands are forwarded to the data FPGA. All responses from the data FPGA are forwarded to the host. Starting commands on both the Control and Data FPGA's without waiting for completion may cause interleaving of response codes.

## **Command Set Overview**

To simplify command processing, commands that are handled in the Control FPGA are assigned in ascending order starting with "G," and commands for the Data FPGA are assigned in descending order starting with "Z."

### **G - Get Camera State**

This command takes no arguments. The camera responds with "G" followed by the current running state of the camera (512 bytes as 1024 hex digits) and a carriage-return. This can be used to synchronize the GUI with the camera at initial connection time. Camera state is sent as hex ASCII unlike the flash page read. This is because the camera state is a single page worth of data, but flash page read is likely to be used many times in a row to read large amounts of data such as the FPGA bitstream.

### **H - Ping Camera for Life**

This command takes no arguments. The camera responds with "H" followed by eight hex digits and a carriage-return. The eight hex digits indicate the value of a free-running 32-bit frame counter. This counter indicates all frames read from the sensor since power on, not necessarily the number of frames sent to the framegrabber. The number of grabbed frames depends on the memory functions such as frame binning, FIFO and circular buffering. Successive pings can be used for a rough estimate of the sensor frame rate.

### **I - Restore Camera State From Flash**

This command takes one argument. The argument can be 01 to 08 and indicates which of the 8 flash storage areas to retrieve the camera settings from. If the flash storage area is uninitialized the camera will send a negative acknowledgement. Otherwise the camera responds with "I" followed by the flash storage area number and a carriage-return. The Restore Camera State From Flash command only restores user settable parameters other than the Data FPGA loads.

### **J - Initialize FPGA From Flash**

This command takes no arguments. The Data FPGA is reloaded from the flash causing a complete re-initialization of the readout logic. If the FPGA fails to load for any reason, the camera will send a negative acknowledge and response code. Otherwise the camera responds with "J" followed by a carriage-return. Failure results when the DONE signal does not go high after download, or if the INIT# signal is reasserted after the bitstream has started to load. Either of these conditions indicates a bitstream error. The Control FPGA does not check the bitstream for correct format or CRC, this is done by the FPGA. To reduce the timeout length if the J command is issued when the flash is uninitialized, the bitstream bit counter only loads the low 24 bits of the bitstream length from the flash header. This still allows up to 128 megabits of bitstream, larger than the flash size. On error a response code is returned. The first byte is the reason which may be 00 01 or 02. 00 indicates bad header flash part type. 01 indicates illegal starting page address. 02 indicates load was attempted but did not complete. If the code is 02, six additional bytes are sent indicating the number of bytes remaining when the operation aborted. This is usually zero, unless the FPGA re-asserted the INIT line.

### **K - Save Current Camera State to Flash**

This command takes one argument. The argument can be 01 to 08 and indicates which of the 8 flash storage areas to store the camera settings in. The camera responds with "K" followed by the flash storage area number and a carriage-return. This command only saves user settable parameters other than the Data FPGA loads (512 bytes total).

## L - Write Flash

This command takes arguments that form the precise command to send to the flash memory. Data length is limited by the flash page buffer size. There is no equivalent write command to "Continuous Array Read". The intent of the command is to allow uploading the FPGA code and to write to flash areas that need to be initialized at the factory. The data is not interpreted by the serial control logic. It is possible to use this command to directly upload camera settings to the flash pages without changing the current camera state.

This is the only Control FPGA command that uses binary encoded arguments (with escape codes for <cr> and \). The exact syntax of the Write Flash command is:

```
'L' <opcode> <addr> <data> <cr>
```

where opcode is one byte, addr is 3 bytes, and data can be from zero to 528 bytes depending on the command. The following opcodes are acceptable to use with this command (there is no error checking so make sure these are the only codes used).

0x84	Buffer 1 write
0x87	Buffer 2 write
0x83	Buffer 1 to main memory program with built-in erase
0x86	Buffer 2 to main memory program with built-in erase
0x88	Buffer 1 to main memory program without built-in erase
0x89	Buffer 2 to main memory program without built-in erase
0x81	Page Erase
0x50	Block Erase
0x82	Main memory page program through buffer 1
0x85	Main memory page program through buffer 2

The following non-write commands are also implemented with Write Flash:

0x53	Main memory Page to Buffer 1 Transfer
0x55	Main memory Page to Buffer 2 Transfer
0x60	Main memory Page to Buffer 1 Compare
0x61	Main memory Page to Buffer 2 Compare
0x58	Auto Page Rewrite through Buffer 1
0x59	Auto Page Rewrite through Buffer 2

The camera responds with "L" followed by a carriage-return. Internally the Control FPGA waits for the flash to become non-busy then starts the flash write operation. It does not wait for the flash to complete (become not busy) after starting the command. Subsequent flash operations check the busy state of the flash memory, allowing overlap of serial communication and flash programming. Theoretically a flash write command without initial wait could give even more overlap by allowing write to a nonbusy buffer while a previous write is ongoing, but in practice the flash page write timing is shorter than the serial transmission time to send the command. At 115,200 baud it takes about 46 mS to send 528 bytes. The flash page erase/write cycle is only 20 mS max.

## M - Read Flash

This command takes arguments that form the precise command to send to the flash memory followed by the command length and data length to be read in bytes. Command length includes any "don't care" bytes required by the flash command selected. Data length is limited by the flash page buffer size for most commands, but can be as large as the entire array if the Continuous Array Read flash command is used. All arguments are ASCII hex. The exact syntax of the Read Flash command is:

```
'M' <opcode> <addr> <cmd_len> <data_len> <cr>
```

where opcode is one byte (2 hex digits), addr is 3 bytes (6 hex digits), cmd\_len is one byte (2 hex digits), and data\_len is four bytes (8 hex digits). Most commands send the <opcode> and <addr> bytes to the

flash and then require some number of "don't care" bytes before starting to read data from the flash. The only exception is the status read command which sends only the command byte and has no additional turn-around delay. When sending the status read command the <addr> bytes are "don't care", but must be included in the command arguments. The following opcodes are acceptable to use with this command (there is no error checking so make sure these are the only codes used).

Opcode	Command	Command Length
D4	Buffer 1 read	5
D6	Buffer 2 read	5
E8	Continuous array read	8
D2	Main Memory Page Read	8
D7	Status Register Read	1

This is the only Control FPGA response that uses binary encoded arguments (with escape codes for <cr> and \).

The camera responds with "M" followed by escaped binary data from the flash and finally a non-escaped carriage-return. Internally the Control FPGA waits for the flash to become non-busy then starts the flash read operation. Thus the Status Register Read command is only useful to check the compare flag, since the busy flag will always be off.

#### **N - Set Camera State**

This command takes a two byte (4 hex digit) offset into the current camera state as defined in Table 4 and from one to 512 bytes (2 to 1024 hex digits) of camera state data. The exact syntax of the Set Camera State command is:

'N' <addr> <data> <cr>

where <addr> is 4 hex digits of offset per Table 4 and <data> is 2 to 1024 hex digits (always a multiple of 2) of data which will be stored sequentially into the current camera state as shown in Table 4. Actual registers affected by the command are updated as the data comes in, except where further synchronization is required such as multibyte values and values that change only between frames. The Control FPGA always responds with "N" followed by a carriage-return even if the affected state is handled by the Data or DDR FPGAs. If the baud rate is changed, the command response happens at the original baud rate.

#### **O - Serial Trigger**

This command takes no arguments. This command has the same effect as pulsing the CC1 line high for the time period between receipt of the 'O' and receipt of the carriage-return. Although this command would normally be used only in edge-triggered mode, it can also be used in external exposure mode for long exposures (minimum exposure is limited by the serial baud rate). To support this mode, any characters between the "O" and the <cr> are ignored. Thus the total command length in characters can be used to determine the exposure time. The camera responds with "O" followed by a carriage-return.

#### **P - Reset and Calibrate Sensor**

This command takes no arguments. It causes the sensor's LRST\_N signal to be pulsed low for two clock cycles. The sensor is reset and performs an auto-calibration. This command is applied by the readout logic between frames. If enabled, the CC4 line will also initiate this command allowing operation without the GUI. The camera responds with "P" followed by a carriage-return.

## **X - Send Initialization Data to Data FPGA**

This command takes any number of bytes (even number of hex digits) of FPGA initialization data. The exact syntax of the Send Initialization Data command is:

'X' <data> <cr>

where <data> is an even number of hex digits of data which can be used for any function required by the Data FPGA. This command is sent by the Control FPGA to the Data FPGA upon power-on initialization and after the 'J' command (Initialize FPGA from Flash). It is also possible for the host to issue this command via the Camera Link serial interface, but be aware that the Data FPGA does not issue a response.

## **Y - Readout Image**

This command takes no arguments. This command is only effective in the memory readout modes. If enabled, the CC2 line will also initiate this command allowing operation without the GUI. The Data FPGA will respond with "Y" followed by a carriage-return.

## **Z - Reset Memory**

This command takes no arguments. This command is only effective in the memory readout modes. If enabled, the CC3 line will also initiate this command allowing operation without the GUI. The Data FPGA will respond with "Z" followed by a carriage-return.

# **13 INTER-FPGA COMMUNICATION**

## **Serial Commands**

Commands from the host are buffered and passed to the Data FPGA on the FPGA\_CTL3 wire. The Data FPGA therefore receives all commands from the host and can act on them accordingly. This allows extensions to be made to the command set for Data FPGA use. To reduce logic in the Data FPGA, each character is buffered and retransmitted at 66 MHz, synchronous to the FPGA\_SYSCCLK. In this way the data FPGA doesn't need to know about baud rate and can start up monitoring commands even before it receives the camera state data.

## **Flash Data**

In addition to forwarding commands from the host, the Control FPGA also uses the FPGA\_CTL3 wire to send camera state data when the Restore Camera State From Flash command is received. It also uses the FPGA\_CTL3 wire to send camera state data and Data FPGA initialization data from flash to the Data FPGA when the Initialize FPGA From Flash command is received. The Control FPGA only responds to the host after it has completed the required data transfer(s). The host must always wait until it receives the response before sending another command.

When transmitting camera state data to the Data FPGA, the Control FPGA uses the same syntax as the Set Camera State command with the starting address set to zero and a data length of 512.

When transmitting Data FPGA initialization data to the Data FPGA, the Control FPGA uses the same syntax as the Set Camera State command with the starting address set to all ones (65,535) and a data length as set in the Page Zero Flash Header. If the header indicates a zero data length, or the initialization data does not start with 3C A5 0F 96, the Control FPGA will not send this command.

## Serial Responses

Normally the data sent to the Data FPGA from the Control FPGA is limited in bandwidth at the source, either by the host serial baud rate or the flash memory read rate. Response data from the Data FPGA, which always gets forwarded to the host, must be sent at the host baud rate. To reduce the redundant logic requirement in the Data FPGA, the Control FPGA uses the FPGA\_DDRCLK wire to send a baud rate pulse train. This signal is high for one cycle of FPGA\_SYSCLK once per serial link bit period. The Data FPGA then uses this as a shift enable for its response transmitter.

Serial response data from the Data FPGA is forwarded to the host directly by ANDing with the serial response data from the Control FPGA. For this reason it is especially important that the host does not send a new command before the previous command response has been received.

## 14 EMBEDDED SOFT PROCESSOR CORE

The Control FPGA includes a "PicoBlaze" 8-bit controller core, programmed in assembly language, that takes care of most low-speed complex operations. This soft processor handles serial protocol, flash memory, FPGA configuration, and power-on initialization. The architecture of this core is Harvard, using a 1K by 18-bit wide instruction memory composed of 5 block RAMs, and an 1K by 8-bit wide data memory composed of four block RAMs. The core was designed and optimized to run in the Virtex 2 series of Xilinx FPGA's and then adapted for the Spartan 2. It therefore is not very fast. In the MV13 it runs at 33 MHz and in the MV40 it runs at 25 MHz (1/2 the pixel clock rate). The instructions run in a fixed 2-cycle period with essentially no pipelining. This simplifies instruction sequence timing calculations.

The PicoBlaze data address space is only 256 bytes. If you look at the processor documents from Xilinx you'll find this called I/O rather than data. In our case the data space connects to 1K bytes of internal block RAM using three additional banking bits implemented as register. Access to the entire 1K of RAM as well as 8 registers is accomplished with some stunt logic. Registers normally appear at addresses 0xf8 through 0xff regardless of the state of the three banking bits. When register access is enabled, writes are shadowed in the block RAM at 0x6f8 through 0x6ff regardless of the state of the three banking bits. Register access can be disabled by reading or writing location 0xf7 and re-enabled by reading or writing locations 0 through 0xf6. In this manner a program can sequentially access the entire RAM by temporarily disabling register access. A special bit, bit 0 of register 0xff toggles each time any read is performed. This allows a simple test of the current state of the register access enable by reading location 0xff twice and comparing the two values.

The high 256 bytes, from 0x700 through 0x7ff, are dual-ported with the I2C slave unit. This provides a simple interface for introducing commands via USB.

### "Fget8" Speed-up logic

Because the PicoBlaze runs relatively slow, there is a small sequencer to speed up FPGA configuration and flash data read. This is kicked off by a write to register 0xfe with bit 6 set. The sequencer runs the flash clock for 8 cycles, and if the fpga CCLK was high, copies the flash DOUT output to the FPGA configuration DIN input cycling the FPGA CCLK 8 cycles as well. The PicoBlaze is responsible for making sure the flash is in the appropriate state when this sequencer is started. Generally it only speeds up the inner loop of FPGA configuration and flash data read, but this is where the largest time is spent. Using the algorithm from previous PicoBlaze designs (Video Combiners) and bit wiggling, the best data rate for FPGA download would be about 1.5 MHz for CCLK. With the "Fget8" speed-up logic the rate should be about 9 MHz. This allows even large FPGA's to be loaded in under 2 seconds.

### PicoBlaze Memory Map

Address	Description
---------	-------------

000 - 1FF	Camera state storage. Sensor registers shadow locations in 000 - 0FF
200 - 3FF	Header storage. Loaded from flash page 00 at power-up.
400 - 6F7	Working memory for host commands.
6F8 - 6FF	Register shadow memory for readback of otherwise write-only bits.
700 - 7FF	I <sup>2</sup> C shared area.

## PicoBlaze Register Map

All of the following registers can only be accessed when the register access enable signal is set. This is accomplished by making any access to locations 00 through 0xf6. Any access to location 0xf7 clears the register access enable signal. Accesses within the 0xf8 through 0xff register area do not affect the state of the register access enable signal.

### 0xF8 Baud Rate Period, low 8 bits and Frame Count.

### 0xF9 Baud Rate Period, high 8 bits and Fget8 data.

The Baud rate is derived by dividing the pixel clock frequency by the number entered here. In the MV40, the pixel clock runs at 66.667 MHz and the default setting for 9600 baud would be 6944. The last value written to the Baud Rate high 8-bits register can be read back from scratchpad location 0x2f9 when register access is disabled. When register access is enabled, reading 0xF9 returns the byte of flash data from the "Fget8" sequencer. Reading 0xF8 returns the frame count in four successive reads. Frame count is latched whenever address 0x00 is read.

### 0xFA Sensor Control. Bits are:

- 7 Read-only "Y" pending. Indicates that a CC2 event has occurred and the PicoBlaze should forward it to the Data FPGA as a "Y" command.
- 6 Read-only "Z" pending. Indicates that a CC3 event has occurred and the PicoBlaze should forward it to the Data FPGA as a "Z" command.
- 5 Reserved.
- 4 Serial trigger. Must be toggled in software to trigger the sensor control logic in response to a serial trigger command.
- 3 Reserved.
- 2 Calibrate sensor. This is a pulsed signal that schedules a calibration when this bit is written to 1. Writing this bit to zero has no effect. Actual calibration may happen much later, since the sensor control logic schedules calibration only between frames.
- 1 Sensor standby. Setting this bit to one places the sensor in low-power standby mode.
- 0 Dark offset enable. Setting this bit allows the sensor to apply internal calibration factors to reduce fixed column noise.

### 0xFB Address Extensions. Bits 3 through 7 are reserved. Other bits are:

- 2,1,0 Banking bits. These bits form the two high address bits to the 1K byte data RAM, except during register access. During register access the high bits are fixed at 6, thus register writes are always shadowed in the 7th of 8 banks, allowing readback of registers that don't have separate read functionality from the shadow RAM. When bank 0 is selected, sensor register write is enabled.

### 0xFC Serial to FPGA and Serial Status.

When written, this register causes a byte of data to be transmitted to the Data FPGA. On reads Bits 0 through 6 are reserved. Other bits are:

- 7 Transmitter to FPGA ready for data.
- 6 Transmitter to FPGA overrun error. Set if write was attempted when the transmitter wasn't ready for data. Cleared when the next data is written when the transmitter is ready for data.

### 0xFD UART to/from Camera Link.

Writing this location sends a byte of serial data to the framegrabber. Reading gets a byte of serial data and acknowledges its receipt. See the status bit descriptions below for more information.

### 0xFE DAC and Flash Control / Flash and UART Status. Bits are:

- 7 Reserved on write, RxData Available on read. When this bit is 1, there is data available to be read from the UART via register 0xFD. The validity of the current data depends on the error bits listed below. This and other receive status bits are cleared when the UART is read.
  - 6 "Fget8" on write, Transmitter Ready on read. Writing this bit to one starts the sequencer for speeding up flash read and fpga configuration write. Writing this bit to zero has no effect. When 1 this bit indicates that the UART is ready to accept a byte of data to transmit to the framegrabber.
  - 5 DAC chip select on write, Receiver framing error on read. DAC chip select is asserted high. This bit is inverted before driving the chip's active low pin. Receiver framing error indicates that a zero was detected in the stop bit position. This condition is cleared when the UART is read.
  - 4 DAC serial clock on write, Receiver overrun error on read. DAC serial clock runs directly to the DAC chips. Receiver overrun indicates that a new character came in when the previous character had not been read. In this case the older data is lost and the data in the UART is the one causing the overrun, i.e. the most recent character received. This condition is cleared when the UART is read.
  - 3 Flash chip select on write, Transmit overrun on read. Flash chip select is asserted high. This bit is inverted before driving the chip's active low pin. Transmit overrun is set if a write was attempted when the transmitter wasn't ready for data. This condition is cleared when data is written to the UART and the transmitter is ready for data.
  - 2 Flash reset on write, transmitter empty on read. Flash reset is asserted high. This bit is inverted before driving the chip's active low pin. Transmitter empty is active when no data transmission is pending. This must be checked before changing the baud rate to ensure any pending characters finish transmitting at the current baud rate.
  - 1 Flash serial clock on write, Flash ready on read. These correspond directly to the flash pin functions.
  - 0 Shared Flash and DAC serial data in on write, Flash data out on read. These correspond directly to the respective pin functions.
- 0xFF FPGA Configuration.** Bit 6 is reserved. Other bits are:
- 7 Read only "Fget8" status. This is high when the sequencer is still running from a prior operation.
  - 5 FPGA Loaded on write, FPGA Done on read. Writing this bit high indicates that the program has finished loading the data FPGA and enables the pinsaver muxes. FPGA Done is the configuration pin function.
  - 4 FPGA M1 on write, FPGA INIT on read. M1 is the configuration mode bit. Other mode bits are pulled up. Writing 1 sets the Data FPGA into slave serial mode for configuration load by the Control FPGA. Writing 0 sets the Data FPGA into JTAG mode for configuration by external Xilinx Parallel Cable. FPGA Init is inverted from the active low configuration /INIT pin.
  - 3 Read only FPGA Data Out. This allows readback functions to be implemented in the future.
  - 2 Write only FPGA Program. This bit is inverted before driving the chip's active low pin. Setting this bit to 1 resets the Data FPGA configuration.
  - 1 FPGA Configuration Clock on write, Configuration Timeout on read. Writing this bit directly affects the FPGA CCLK pin except when the "Fget8" sequencer is running. In that case this bit enables Flash to FPGA data transfer if it is 1. Running "Fget8" while this bit is 0 will not cause any changes to the FPGA DIN or CCLK signals. Configuration Timeout goes high if there have been 16,777,216 CCLK clocks since Program was last asserted. This can be used as a simple timeout mechanism for a program that only checks the Done and Init status during FPGA configuration.
  - 0 FPGA Data In on write, Access Test Bit on read. During FPGA configuration this is the serial data to the Data FPGA. After configuration it could be used as an additional control line to the Data FPGA. The Access Test Bit toggles on any port read. Thus reading this register twice in a row will always provide different data. This can be used to test whether the register access enable signal is active.



## Sensor Register Map

All of the following registers can only be accessed when the sensor register write enable signal is set. This is bit 2 of register 0xfb. Writing these registers will also cause the values to be shadowed in the scratchpad RAM at whatever page is currently indicated by bits 1 and 0 of register 0xfb. Thus although the registers themselves are write-only, there is a copy in the scratchpad RAM immediately after writing them. The scratchpad copy can obviously be overwritten without affecting the sensor registers if the sensor register write enable bit is off. The layout of these registers intentionally matches the flash state storage.

**0x24 Start Pixel, low 8 bits.**

**0x25 Start Pixel, high 8 bits.** This is the leftmost pixel in the ROI. Pixels are numbered from 0 to the sensor width - 1. This is different from the old ROI settings which worked in pixel clocks. The sensor control logic handles the conversion from pixels to clocks. If the starting pixel is not a multiple of the sensor readout width (10 for MV13 or 16 for MV40) there will be some prescan pixels in the resulting image. In the future we may want to add logic in the data FPGA to mask out the prescan pixels.

**0x26 End Pixel, low 8 bits.**

**0x27 End Pixel, high 8 bits.** This is the rightmost pixel in the ROI. Pixels are numbered from 0 to the sensor width - 1. This is different from the old ROI settings which worked in pixel clocks. The sensor control logic handles the conversion from pixels to clocks. If the ending pixel is not one less than a multiple of the sensor readout width (10 for MV13 or 16 for MV40) there will be some postscan pixels in the resulting image. In the future we may want to add logic in the data FPGA to mask out the postscan pixels.

**0x28 Start Line, low 8 bits.**

**0x29 Start Line, high 8 bits.** This is the top line in the ROI. Lines are numbered from 0 to the sensor height - 1. Only lines in the ROI are scanned.

**0x2A End Line, low 8 bits.**

**0x2B End Line, high 8 bits.** This is the bottom line in the ROI. Lines are numbered from 0 to the sensor height - 1. Only lines in the ROI are scanned.

**0x2C Line Period, low 8 bits.**

**0x2D Line Period, high 8 bits.** This is the line readout time in pixel clocks - 1. i.e. the actual period is one clock more than this number.

**0x2E Exposure Period, bits <7:0>**

**0x2F Exposure Period, bits <15:8>**

**0x30 Exposure Period, bits <23:16>**

**0x31 Exposure Period, bits <31:24>** This is the desired exposure time in pixel clocks. Actual exposure time may vary, especially if readout overlaps exposure.

**0x32 Frame Period, bits <7:0>**

**0x33 Frame Period, bits <15:8>**

**0x34 Frame Period, bits <23:16>**

**0x35 Frame Period, bits <31:24>** This is the desired frame readout period in pixel clocks - 1. i.e. the actual period is one clock more than this number.

**0x36 Delay Period, bits <7:0>**

**0x37 Delay Period, bits <15:8>**

**0x38 Delay Period, bits <23:16>**

**0x39 Delay Period, bits <31:24>** This is the desired delay from trigger to exposure in edge-triggered modes. It is not used in free-running or external exposure modes. Actual exposure delay may vary, especially if readout overlaps exposure.

**0x41 Trigger Mode.** Bit 7 is reserved. Other bits are:

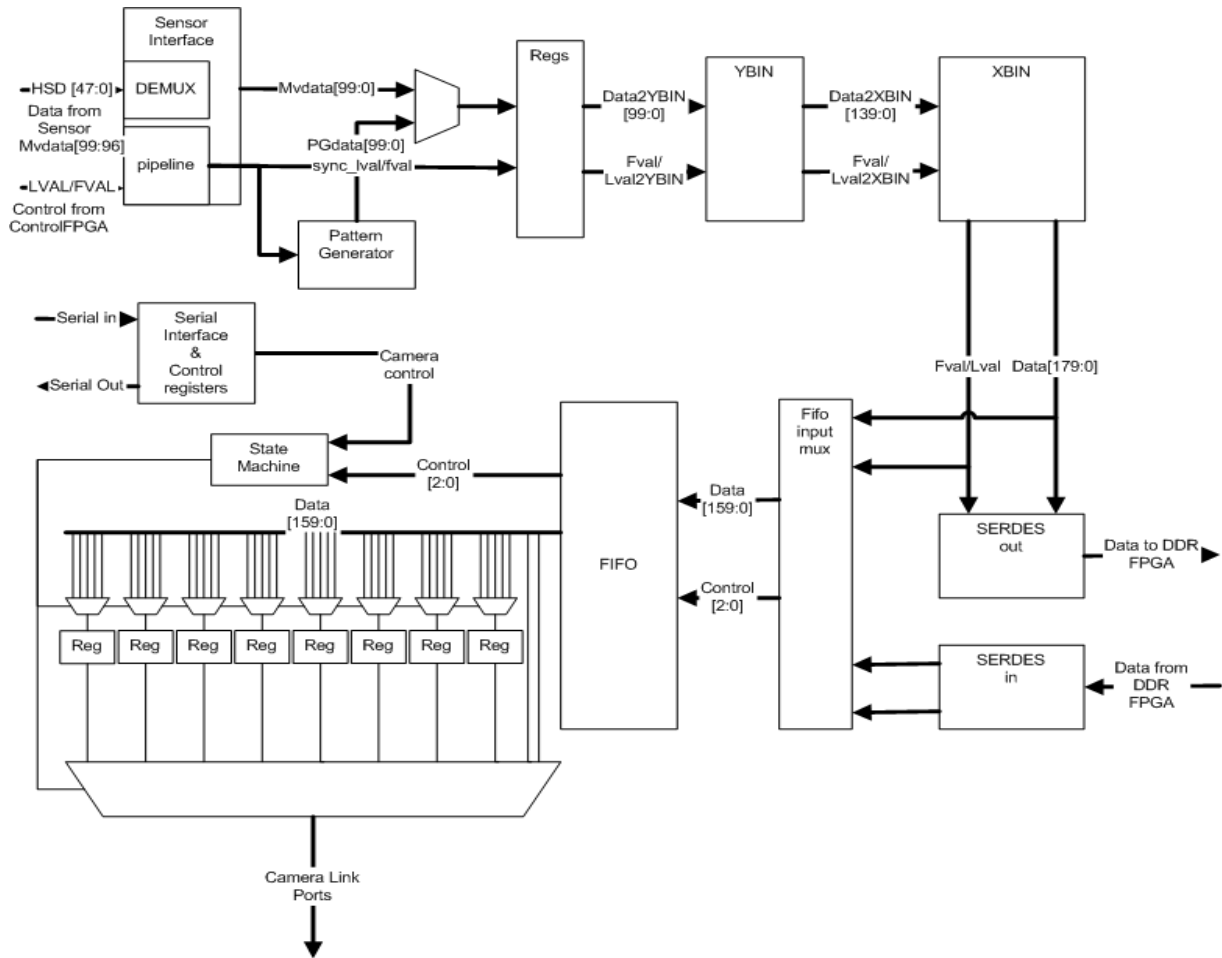
**6** Invert TTL Trigger. Setting this bit to 1 indicates the TTL trigger input on P2 is active low level, or falling edge.

**5** Enable TTL trigger. Set this bit to 1 to enable TTL trigger input on P2. Clear to 0 to disable TTL trigger input.

- 4 Synchronous Exposure. This bit only affects free-run and multi-frame edge-triggered modes. When this bit is set, the readout timing is synchronized to the exposure.
  - 3 Invert Camera Link CC1 Trigger. Setting this bit to 1 indicates the CC1 trigger input on Camera Link is active low level, or falling edge.
  - 2 Enable Camera Link CC1 trigger. Set this bit to 1 to enable CC1 trigger input on Camera Link. Clear to 0 to disable CC1 trigger input.
  - 1:0 Trigger Mode.
    - 0 Free-running mode.
    - 1 Multi-frame edge-triggered mode.
    - 2 Single edge-triggered mode.
    - 3 External exposure mode.
- 0x43 Trigger Count.** Total number of frames to capture in multi-frame edge-triggered mode. 0 to 255 program 0 to 255 frames.
- 0x44 CC Mode.** Enable and active edge of CC2, CC3, and CC4 command inputs. Bits 7, 6 are reserved. Other bits are:
- 5 Invert CC4. Setting this bit to 1 indicates the CC4 input is active on falling edge. Setting this bit to 0 indicates the CC4 input is active on rising edge.
  - 4 Enable Camera Link CC4. Set this bit to 1 to enable CC4 Sensor Calibration. When enabled, the active edge of CC4 causes the sensor to be reset and calibrated at the next appropriate time.
  - 3 Invert CC3. Setting this bit to 1 indicates the CC3 input is active on falling edge. Setting this bit to 0 indicates the CC3 input is active on rising edge.
  - 2 Enable Camera Link CC3. Set this bit to 1 to enable CC3 Reset Memory. When enabled, the active edge of CC3 causes the memory to be reset.
  - 1 Invert CC2. Setting this bit to 1 indicates the CC2 input is active on falling edge. Setting this bit to 0 indicates the CC2 input is active on rising edge.
  - 0 Enable Camera Link CC2. Set this bit to 1 to enable CC2 Readout Image. When enabled, the active edge of CC2 causes the current memory image to be read out.

## 15 DATA FPGA

The system level block diagram for the FPGA is shown below.



## 15.1 Video Data path overview

The DataFpga accepts video data from the sensor and formats it for transmission over the Camera Link outputs or USB interface. The ControlFpga controls the sensor data output and sends Lval (line valid) and Fval (frame valid) signals to the DataFpga. The data from the sensor is valid when Lval and Fval are asserted. Data progresses through to the FIFO (or output to the DDR Fpga in memory mode) at the pixel rate of the sensor and is the width of the sensor data (100 or 160 bits wide).

The video data from the sensor is sent to the DataFpga over the HSD[47:0] and Mvdata[99:96] signal lines. The MV13 sensor clocks out ten 10bit pixels (100 bits total) each clock (at 66Mhz) and the MV40 sensor clocks out sixteen 10bit pixels (160 bits total) each clock (60Mhz). These data lines are multiplexed to the DataFpga over the HSD (high speed data) signal lines at 2x the pixel rate in order to reduce the I/O count needed by this Fpga. The DataFpga demultiplexes this data and synchronizes it with the non multiplexed data from the sensor and the control signals from the ControlFpga.

The video data is then sent into the vertical binning module. This module accumulates the programmed number of rows of video data and then forwards the binned row of data to the horizontal binning module. If vertical binning is active then the frame height is reduced by the binning factor. The vertical binner produces 140 bits of video data as ten 14bit pixels. The horizontal binning module adds together the programmed number of pixels and then forwards the reduced row of data to the FIFO input mux. The horizontal binner produces 180 bits of data as ten 18bit pixels. The FIFO input mux selects either the upper or lower 16bits from each 18 bit pixel based upon the binning multiplier mode selection (also known as Sensor Resolution). The video data is then stored in a large FIFO along with the state of Lval/Fval for each block of ten (or 16) pixels.

The section of the FPGA that reads the video data from the FIFO operates at the specified Camera Link clock rate. This rate can be selected to be 33, 42.5, 66 or 85Mhz. The output state machine controls the reading of the FIFO and outputting the data to the camera links based upon the selected number of Camera Link taps and the pixel width selected. There are a multitude of Camera Link modes supported by this camera, from basic one tap mode up to ten tap 80bit mode.

The output stage over the Camera Link does not use the Dval signal as defined within the Camera Link specification. This is done to insure compatibility with a variety of framegrabbers. In order to do this, the output state machine waits until enough data has been accumulated in the FIFO in order to not under-run the FIFO while the line of video data is being output. This is only a concern at higher output modes like 8 to 10 tap modes running at 66 or 85Mhz Camera Link output clock speeds. This does not affect the maximum frame rate achievable by this camera, it just insures that once Lval is asserted out the Camera Link the whole line of data will be output without gaps.

## **15.2 Operating mode control overview**

The operating state of the DataFpga is set by the ControlFpga or by the host via the serial interface. The ControlFpga can load stored camera configurations from the system flash memory at powerup or when directed to load different configurations by the host.

### 15.3 Camera State

Internal to the ControlFpga all the camera state is saved in a Block RAM. Copies of the current state can be saved to the flash or uploaded to the host. The current state can also be retrieved from flash or changed by the host. Only the host has random access to the camera state and this only when setting state. Reading back the camera state always sends the entire state to the host. Table 4 shows the layout of the camera state memory. Except for sensor reference voltages, multibyte values are little endian. The values shown in bold apply to the DataFpga and the values shown in grey apply to the ControlFpga.

Byte Offset (decimal)	Bytes	Description
0	4	C3, 5A, F0, 69 for detecting uninitialized buffers
4	2	VIn2: 14 D9
6	2	Vref1: 14 D9
8	2	Vtest: 20 00
10	2	Vref2: 23 E0
12	2	Vbias1: 30 00
14	2	Vref3: 32 E8
16	2	Vbias2: 40 00
18	2	Vref4: 41 36
20	2	Vbias3: 50 00
22	2	VIn1: 54 D9
24	2	Vbias4: 60 00
26	2	VIp: 64 D9
28	2	Vunused1: 70 00
30	2	Vclamp3: 70 00
32	2	Vunused2: 80 00
34	2	Vrstpix: 8D 17
36	2	ROI Start Pixel
38	2	ROI End Pixel
40	2	ROI Start Line
42	2	ROI End Line
44	2	<b>Line Period in Pixel Clocks</b>
46	4	Exposure Time in Pixel Clocks
50	4	Frame Period in Pixel Clocks
54	4	Exposure Delay in Pixel Clocks
58	2	Serial Link Bit Period in Pixel Clocks
60	1	<b>Camera Link Readout Mode</b>
61	1	<b>Camera Link Clock Frequency</b>
62	1	<b>Binning control (horizontal and vertical binning multipliers)</b>
63	1	<b>Memory Options</b>
64	1	<b>Binning Multiplier</b>
65	2	Trigger Mode (2nd Byte is Data FPGA dependent)
67	1	Frame count for Multi-Trigger mode
68	1	CC Mode: CC2, CC3, CC4 enable and edge select
69	59	Reserved for Control FPGA / base system extension
128	384	<b>Available for Data FPGA functionality extensions</b>

Table 4 - Camera State Memory Layout

Whenever the state memory is updated from host or flash, the actual internal registers that implement the camera state change, too. Some of these are located in the Control FPGA and some in the Data FPGA. The Control FPGA forwards state data to the Data FPGA whenever it is updated.

## 15.4 DataFpga State controls

The following section details the modes selectable within the DataFpga.

### 15.4.1 Camera Link Readout Mode

This byte controls the operating mode of the Camera Link Interface. The table below shows the 'ports' assigned to each camera link connector output for the selectable modes. The Camera Link specification defines three 8bit ports A,B,C per each Camera Link channel which consists of 5 wires. In the table below CL1 is the Camera Link channel on connector J1. CL2 and CL3 are Camera Link channels on connector J2.

Mode hex value	Camera Link Format MSB ----- LSB	Read Out Mode
0x00	CL1_A[7:0]	Single tap 8 bits (Basic)
0x01	CL1_B[1:0],CL1_A[7:0]	Single tap 10 bits (Basic) (default mode)
0x02	CL1_B[3:0],CL1_A[7:0]	Single tap 12-bits (Basic)
0x03	CL1_B[7:0],CL1_A[7:0]	Single tap 16-bits (Basic)
0x04	CL1_A[7:0] even pixels (0,2,...) CL1_B[7:0] odd pixels (1,3,...)	Two Taps 8 bits (Basic)
0x05	CL1_B[1:0],CL1_A[7:0] even CL1_B[5:4],CL1_C[7:0] odd	Two Taps 10 bits (Basic)
0x06	CL1_B[3:0],CL1_A[7:0] even CL1_B[7:4],CL1_C[7:0] odd	Two Taps 12 bits (Basic)
0x0A	CL1_A[7:0] pixels 4*n+0 CL1_B[7:0] pixels 4*n+1 CL1_C[7:0] pixels 4*n+2 CL2_A[7:0] pixels 4*n+3	Four Taps 8 bits (Medium)
0x0B	CL1_B[1:0],CL1_A[7:0] 4*n+0 CL1_B[5:4],CL1_C[7:0] 4*n+1 CL2_C[1:0],CL2_B[7:0] 4*n+2 CL2_C[4:5],CL2_A[7:0] 4*n+3	Four Taps 10 bits (Medium)
0x0C	CL1_B[3:0],CL1_A[7:0] 4*n+0 CL1_B[7:4],CL1_C[7:0] 4*n+1 CL2_C[3:0],CL2_B[7:0] 4*n+2 CL2_C[7:4],CL2_A[7:0] 4*n+3	Four Taps 12 bits (Medium)
0x0D	CL1_B[7:0],CL1_A[7:0] 4*n+0 CL2_A[7:0],CL1_C[7:0] 4*n+1 CL2_C[7:0],CL2_B[7:0] 4*n+2 CL3_B[7:0],CL3_A[7:0] 4*n+3	Four Taps 16 bits (Medium)
0x12	CL1_A[7:0] 8*n+0 CL1_B[7:0] 8*n+1 CL1_C[7:0] 8*n+2 CL2_A[7:0] 8*n+3 CL2_B[7:0] 8*n+4 CL2_C[7:0] 8*n+5 CL3_A[7:0] 8*n+6 CL3_B[7:0] 8*n+7	Eight Taps 8 bits (Full)
0x13	(pin assignment below)	Eight Taps 10 bits (Full)
0x14	(pin assignment below)	<b><u>TEN TAPS 8 BITS</u></b>

254	NA	Use Serial Port
255	NA	Use USB

### 8 Tap 10 bitmode

This mode uses all of the RX/TX bits of the camera link interface so it needs to be described in terms of the actual RX/TX bits versus the Camera Link Specification which uses virtual Ports A-J. The port bits in the following table refer to the 8-10bit pixels A-H.

Port Bit	Tx bit	Port Bit	Tx bit	Port Bit	Tx bit
A0	Tx0	C6	Tx0	F3	Tx0
A1	Tx1	C7	Tx1	F4	Tx1
A2	Tx2	C8	Tx2	F5	Tx2
A3	Tx3	C9	Tx3	F6	Tx3
A4	Tx4	D0	Tx4	F7	Tx4
A5	Tx5	D1	Tx5	F8	Tx5
A6	Tx6	D2	Tx6	F9	Tx6
A7	Tx7	D3	Tx7	G0	Tx7
A8	Tx8	D4	Tx8	G1	Tx8
A9	Tx9	D5	Tx9	G2	Tx9
B0	Tx10	D6	Tx10	G3	Tx10
B1	Tx11	D7	Tx11	G4	Tx11
B2	Tx12	D8	Tx12	G5	Tx12
B3	Tx13	D9	Tx13	G6	Tx13
B4	Tx14	E0	Tx14	G7	Tx14
B5	Tx15	E1	Tx15	G8	Tx15
B6	Tx16	E2	Tx16	G9	Tx16
B7	Tx17	E3	Tx17	H0	Tx17
B8	Tx18	E4	Tx18	H1	Tx18
B9	Tx19	E5	Tx19	H2	Tx19
C0	Tx20	E6	Tx20	H3	Tx20
C1	Tx21	E7	Tx21	H4	Tx21
C2	Tx22	E8	Tx22	H5	Tx22
C3	Tx23	E9	Tx23	H6	Tx23
LVAL	Tx24	F0	Tx24	H7	Tx24
FVAL	Tx25	F1	Tx25	H8	Tx25
C4	Tx26	F2	Tx26	H9	Tx26
C5	Tx27	LVAL	Tx27	LVAL	Tx27

### 10 Tap 8 bit mode

This mode (Basler A501k mode) uses all of the RX/TX bits of the camera link interface so it needs to be described in terms of the actual RX/TX bits versus the Camera Link Specification which uses virtual Ports A-J. The port bits in the following table refer to the 10 8bit pixels A-K.

Port Bit	Tx bit	Port Bit	Tx bit	Port Bit	Tx bit
A0	Tx0	D2	Tx0	G5	Tx0
A1	Tx1	D3	Tx1	G6	Tx1

A2	Tx2	D4	Tx2	G7	Tx2
A3	Tx3	D5	Tx3	H0	Tx3
A4	Tx4	D6	Tx4	H1	Tx4
A5	Tx5	D7	Tx5	H2	Tx5
A6	Tx6	E0	Tx6	H3	Tx6
A7	Tx7	E1	Tx7	H4	Tx7
B0	Tx8	E2	Tx8	H5	Tx8
B1	Tx9	E3	Tx9	H6	Tx9
B2	Tx10	E4	Tx10	H7	Tx10
B3	Tx11	E5	Tx11	J0	Tx11
B4	Tx12	E6	Tx12	J1	Tx12
B5	Tx13	E7	Tx13	J2	Tx13
B6	Tx14	F0	Tx14	J3	Tx14
B7	Tx15	F1	Tx15	J4	Tx15
C0	Tx16	F2	Tx16	J5	Tx16
C1	Tx17	F3	Tx17	J6	Tx17
C2	Tx18	F4	Tx18	J7	Tx18
C3	Tx19	F5	Tx19	K0	Tx19
C4	Tx20	F6	Tx20	K1	Tx20
C5	Tx21	F7	Tx21	K2	Tx21
C6	Tx22	G0	Tx22	K3	Tx22
C7	Tx23	G1	Tx23	K4	Tx23
LVAL	Tx24	G2	Tx24	K5	Tx24
FVAL	Tx25	G3	Tx25	K6	Tx25
D0	Tx26	G4	Tx26	K7	Tx26
D1	Tx27	LVAL	Tx27	LVAL	Tx27

#### 15.4.2 Camera Link Clock Frequency setting

This byte controls the Camera link output clock rate.

FC13 Camera Link Clock Frequency	
0x00	33MHz
0x01	42.5MHz
0x02	66MHz
0x03	85MHz

FC40 Camera Link Clock Frequency	
0x00	30MHz
0x01	42.5MHz
0x02	60MHz
0x03	85MHz

#### 15.4.3 Binning Control setting

This byte controls which 16 bits are selected from the 18 bit binned pixel value.

Binning Control byte	
Bits [3:0]	Horizontal binning setting with rates selectable in the range of 1 to 14.
Bits [7:0]	Vertical binning setting with rates selectable in the range of 1 to 12.



#### 15.4.4 Binning Multiplier setting

This byte controls which 16 bits are selected from the 18 bit binned pixel value.

Binning Multiplier byte	
0x00	Bits [15:0] selected from the 18 bit binned pixel value.
0x01	Bits [17:2] selected from the 18 bit binned pixel value.

The actual Frame rate of the camera is determined by settings within the ControlFpga and the DataFpga.

### 15.5 Frame Rates

The ControlFpga settings determine the data rate from the sensor by adjusting the ROI (Region of Interest), exposure rate and exposure delay. The DataFpga settings determine the binning settings (which can increase frame rate as there is less data to send over the Camera Link Interface), the Camera link mode and clock speed. To get the fastest frame rate possible for a given ROI you can calculate the minimum required Line Period (in pixels clocks) by the following formula.

$$\text{Min. Line Period} = \frac{(\text{Time\_required\_over\_Camera\_Link} - \text{Time\_to\_gather\_data\_from\_sensor})}{1/66\text{Mhz}} - 132 \text{ (clocks per line)}$$

The time required to send each row of data over the Camera Link is calculated by:

$$\text{Time\_required\_over\_Camera\_Link} = \left( \frac{\text{Num\_Data\_pixels\_per\_row}}{\text{Num\_Taps}} \right) * 1/\text{Selected\_Clock\_rate}$$

The time required to gather each row of data from the sensor is calculated by:

$$\text{Time\_to\_gather\_data\_from\_sensor} = 132 \text{ (clocks per row min)} * 1/66\text{Mhz} = 2\text{us}$$

$$\text{Num\_Data\_pixels\_per\_row} =$$

$$\left( \frac{(\# \text{ of cols in ROI} * 10\text{pixels/column})}{\text{Horiz\_binning}} \right) / \text{Vertical\_binning}$$

So for example, when running full resolution (1280 pixels per row), no binning, in 1Tap mode at 33Mhz clock rate the calculation is:

$$\text{Time\_required\_over\_Camera\_Link} - 2\text{us} / 1/66\text{Mhz} - 132 \text{ (clocks per line)}$$

$$38.7\text{us} - 2\text{us} / 1/66\text{Mhz} = 2422 - 132 = 2290$$

Thus the Minimum line period needs to be set to 2290, which adds that many clocks per line read from the sensor, which results in a frame rate of  $1/(1024 \text{ rows} * 38.7\text{us per row})$  or about 25 full resolution frames per second.

If we up the clock rate over the Camera Link to 85Mhz and switch to 8Tap mode then the limiting factor becomes the time to gather the data from the sensor:

$$\text{Time\_required\_over\_Camera\_Link} = 1280/8 * 1/85\text{Mhz} = 1.88\text{uS}$$

The time duration of 1.88uS to output the line is less than the time required to gather the sensor data of 2uS so we do not need to extend the line length and the Minimum Line period can be set to 0.

## 15.6 Memory Operation.

The FC40 DataFpga connects to the DdrFpga via a high speed SERDES link. This link is 13 bits wide and is able to transmit 104 bits of data to the DdrFpga at the sensor clock speed (66Mhz). This data path

## 15.7 Technical Details.

consists of 100 bits of data (10, 10bit pixels), 2 bits of control (Lval, Fval), and 2 spare bits.

The following diagrams show simplified state diagrams used to design the Camera Link output stage of the DataFpga.

The data from the fifos are fed to 5-to-1 muxes on order to limit the fanout of the data from the FIFO. There are 8 muxs in order to support 8 Tap mode. Muxs 1,3,5,7 multiplex between pixels 1,3,5,7,9 and Muxs 2,4,6,8 multiplex between pixels 2,4,6,8,T. There are 4 mux select buses which control Muxs 1-2, 3-4, 5-6, 7-8.

The ordering of the pixels output each clock vary per the number of taps:

# of Taps	Order of the pixels output at each clock cycle	Pixels read during the current clock cycle to be available for on subsequent clock cycles
1 Tap	1 2 3 4 5 6 7 8 9 T	1 2 3 4 5 6 7 8 9 T (whole FIFO preread)          1 2 3 4 5 6 7 8          9 T
2 Taps	1 2 3 4 5 6 7 8 9 T	1 2 3 4 5 6 7 8 9 T (whole FIFO preread)          1 2 3 4 5 6 7 8          9 T
4 Taps	1234 5678 9T12 3456 789T	1 2 3 4 5 6 7 8 9 T (whole FIFO preread)          1 2 3 4 5 6 7 8          9 T          1 2 3 4 5 6          7 8 9 T
8 Taps	1 2 3 4 5 6 7 8 9 T 1 2 3 4 5 6 7 8 9 T 1 2 3 4 5 6 7 8 9 T 1 2	1 2 3 4 5 6 7 8 9 T (whole FIFO preread)          1 2 3 4 5 6          1 2 3 4 7 8 9 T          1 2 5 6 7 8 9 T          3 4 5 6 7 8 9 T

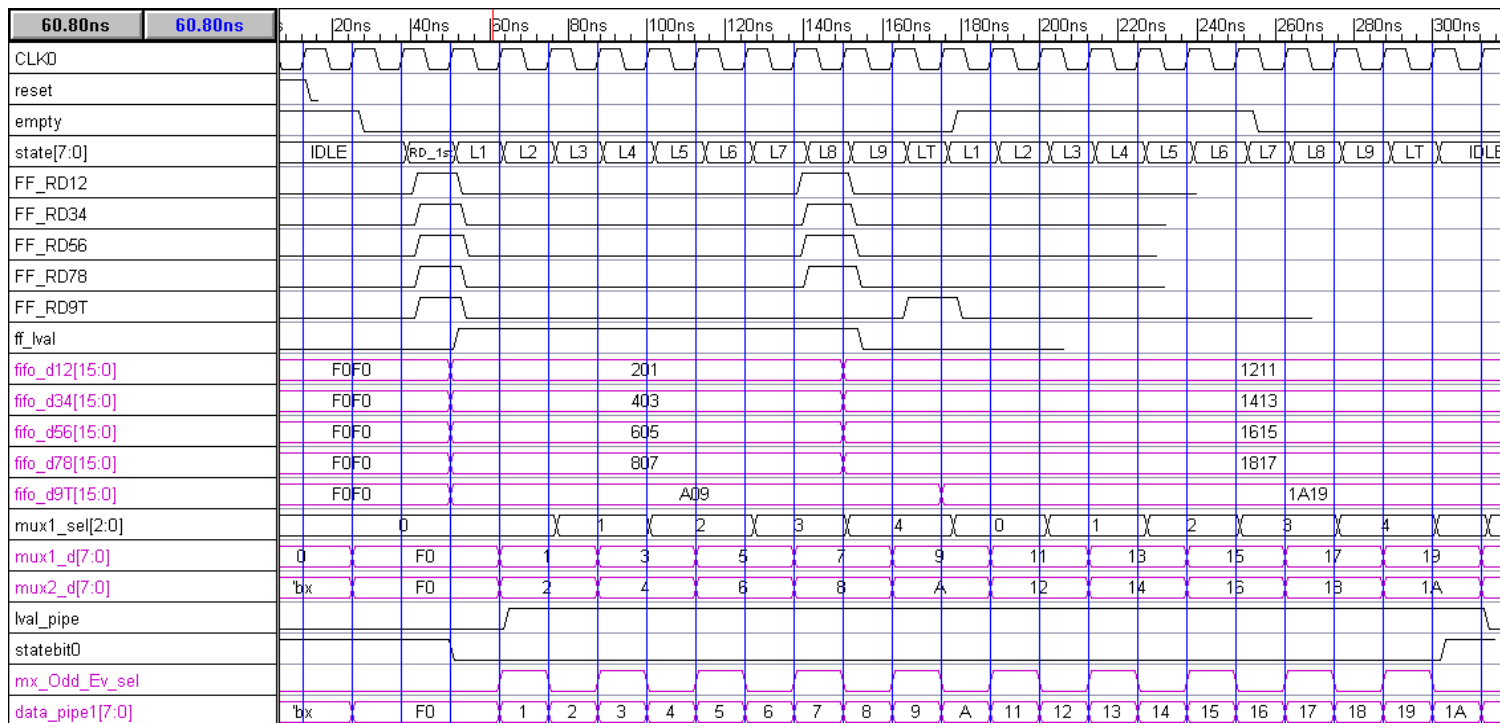
	3 4 5 6 7 8 9 T	1 2 3 4 5 6 7 8 9 T
10 Taps	1 2 3 4 5 6 7 8 9 T	1 2 3 4 5 6 7 8 9 T (whole FIFO pre-read)
	1 2 3 4 5 6 7 8 9 T	1 2 3 4 5 6 7 8 9 T

The following notes apply for these timing diagrams

- CLK0 is the camera link output clock
- Empty is the signal from the FIFO that starts the state machine operation in these simplified state diagrams. The actual verilog code used a combination of Empty and other conditions to insure that the FIFO did not under-run.
- State[7:0] signal uses mnemonics to indicate the state machine states. The state machine starts in Idle, changes to RD\_1<sup>st</sup> (first fifo read) and then loops reading the FIFO and outputting data until the Lval signal from the FIFO deasserts.
- The FIFO was split into 5 sperate FIFO's each with their own read control signals. The five FIFOs contain pixels 1-2, 3-4, 5-6, 7-8, 9-10. The data from the FIFOs was shown here as a byte per pixel, so signal fifo\_d12[15:0] represents pixel2 on bits [15:8] and pixel1 on bits [7:0]
- Signal FF\_RDxx are the FIFO read signals.
- Signal ff\_lval is the Lval status out of the FIFO.
- Signals fifo\_dxx[15:0] are the pixels from the FIFO.
- Mux1\_sel[2:0] controls the mux selects for muxs 1,2. Mux2\_sel controls muxs 3,4,etc. Thus a mux select index of 0x0 for muxs1,2 selects the first pixel from the two muxs which are pixels 1 and 2.
- Signals MuxX\_d show the pixels selected from the mux

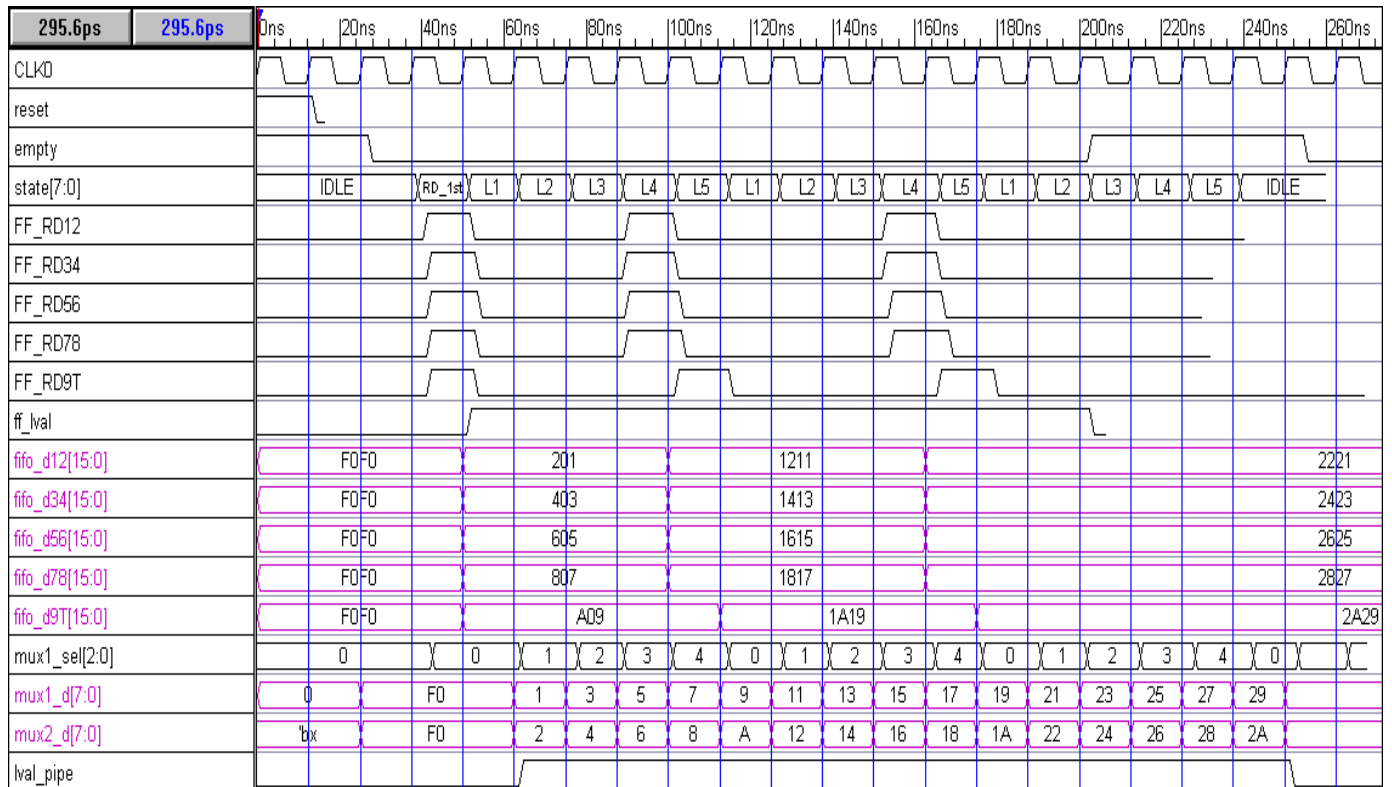
### 15.8 1 Tap mode

When the FIFO deasserts empty the state machine starts with a pre-read of the whole FIFO during which pixels 1-Ten are read. The state machine then continues in a 10 cycle loop until the end of the line is detected (ff\_lval deasserted). The outputs from muxs 1,2 are multiplexed again onto the single output tap (data\_pipe1)



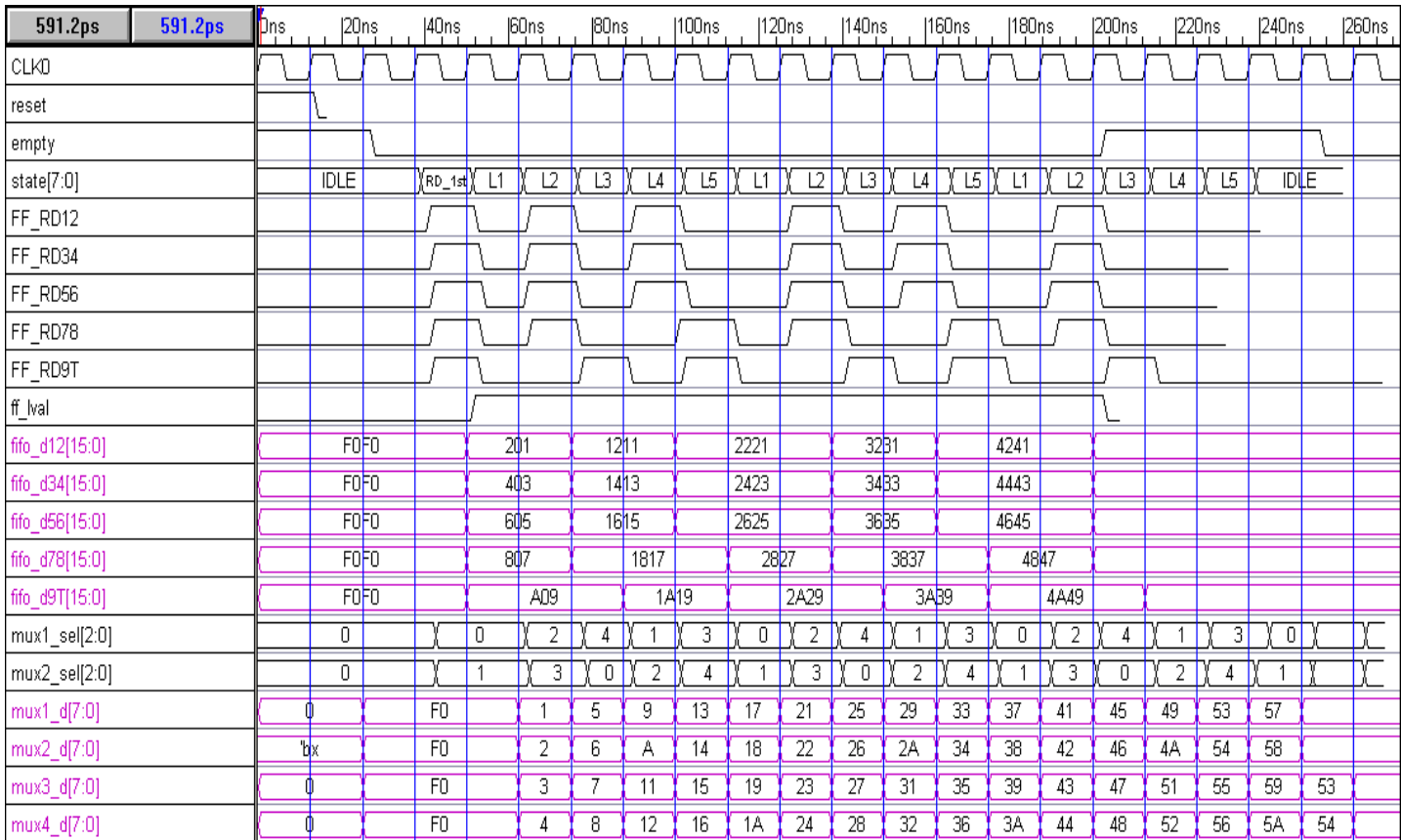
## 15.9 2 Tap mode

When the FIFO deasserts empty the state machine starts with a pre-read of the whole FIFO during which pixels 1-Ten are read. The state machine then continues in a 5 cycle loop until the end of the line is detected (ff\_lval deasserted). The output from muxs 1,2 drive the two taps.



## 15.10 4 Tap mode

When the FIFO deasserts empty the state machine starts with a pre-read of the whole FIFO during which pixels 1-Ten are read. The state machine then continues in a 5 cycle loop until the end of the line is detected (ff\_lval deasserted). The output from muxs 1,2,3,4 drive the four taps.



### 15.11 8 Tap mode

When the FIFO deasserts empty the state machine starts with a pre-read of the whole FIFO during which pixels 1-Ten are read. The state machine then continues in a 5 cycle loop until the end of the line is detected (ff\_lval deasserted). The output from muxs 1,2,3,4,5,6,7,8 drive the eight taps.



framegrabber is being used. A camera control program comes with the camera, which uses the serial port supported by the framegrabber, if it supports the standard Camera Link DLL, or may be used with the USB port in situations where the camera link serial port is not available, or too slow. In order to use this option you must be using an operating system that supports USB, and have the correct drivers installed. Currently this is supported under Windows and Linux only.

The USB port may also be used to obtain image data from the camera, you may select full images or an ROI to reduce the image rate. The camera supports USB 2.0, which provides up to about 40 MBytes/sec transfer rates. You will probably get less, it depends on your computer; typically the transfer rate is more than 30 Mbytes/sec.

## **17 CAMERA CONTROL PROGRAM**

The Camera Control Program provides a GUI interface that allows you to send and receive control data and images from the camera. See read me files under camera control program software.

## **18 APPLICATION ENVIRONMENTS**

In this section we will discuss several operating environments to give you an idea of what you need to operate the camera with a PC.

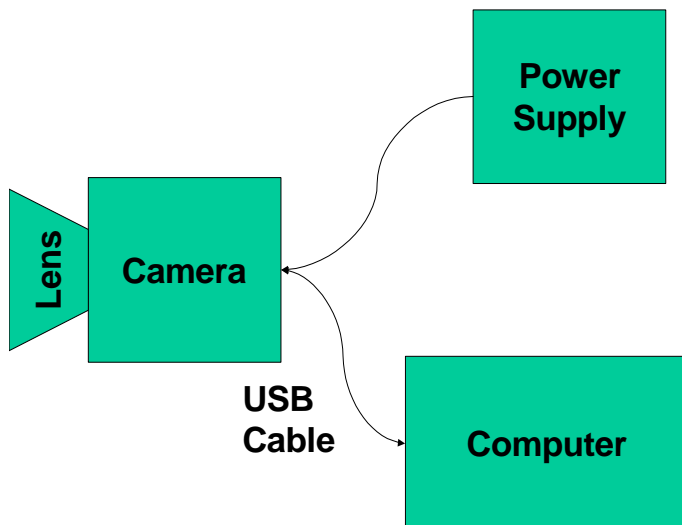
### **18.1 USB operation**

In this mode of operation needs the following items:

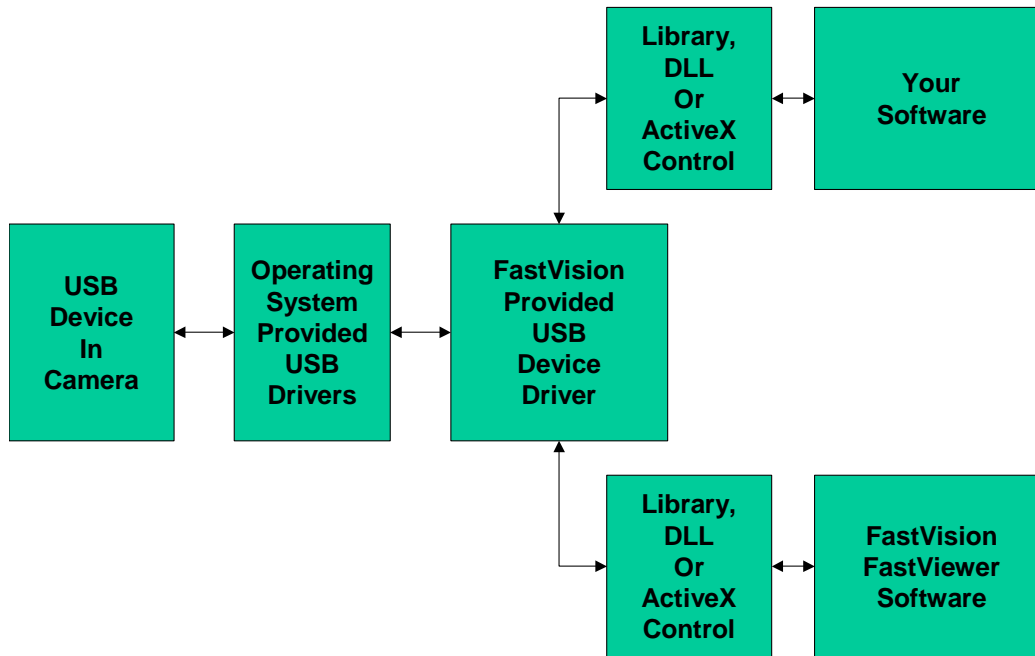
- **The camera and a lens**
- **A power supply for the camera.**
- **A USB cable**
- **FastVision supplied USB software for your OS.**

All of the above can be supplied to you by FastVision. The software (item #4) provides two forms of access to the camera, via a library (DLL, or ActiveX control under Windows, linkable library under other operating systems), and via the supplied application called FastViewer.

#### **18.1.1 The hardware connections are:**



### 18.1.2 The software connections are:

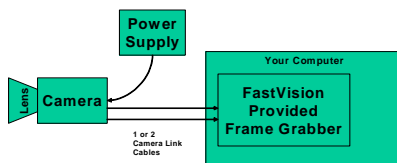


## 18.2 Using a FastVision supplied framegrabber

In this mode of operation you will need the following items:

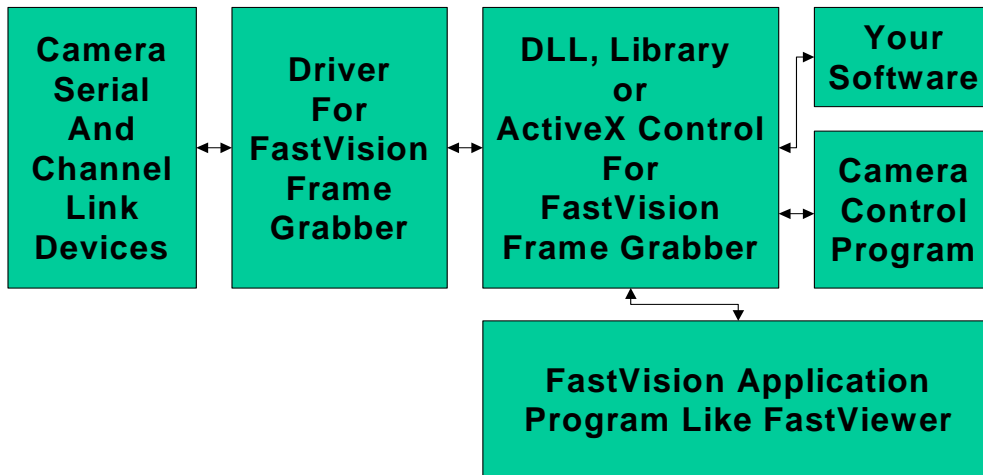
- **The camera and a lens**
- **A power supply for the camera.**
- **One or more camera link cables.**
- **FastVision supplied framegrabber**
- **FastVision supplied framegrabber software.**
- All of this can be supplied by FastVision.

### 18.2.1 The hardware connections are:



### 18.2.2 The software connections are:



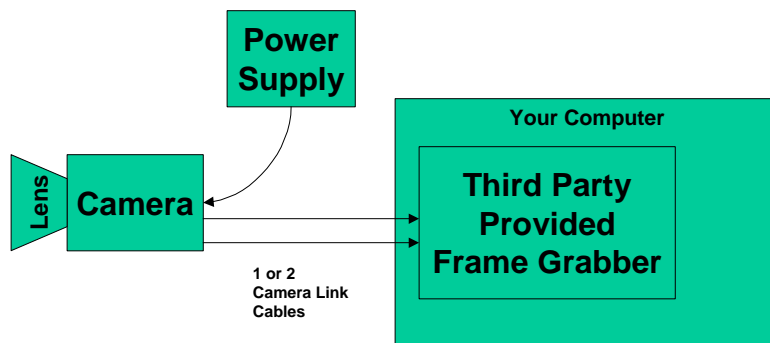


### 18.3 Using a third-party framegrabber

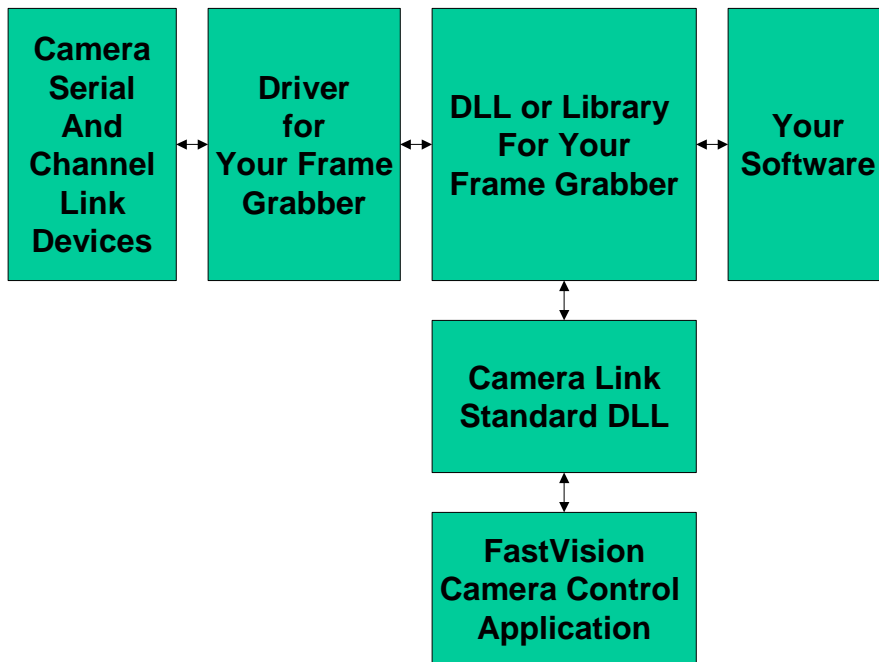
In this mode of operation you will need the following items:

- The camera and a lens
- A power supply for the camera.
- One or more Camera Link cables for the camera.
- Third-party framegrabber
- AIA compliant software provided by FG supplier
- FastVision supplied camera control program

#### 18.3.1 The hardware connections are:



### 18.3.2 The software connections are:



This will only work if your framegrabber supports the Camera Link standard serial interface protocol, and your framegrabber vendor has provided the needed DLL. Otherwise you will have to send commands to the camera in the way specified by your framegrabber vendor.

## **19 TROUBLESHOOTING**

There are several things you can try before you call FastVision Technical Support for help.

- \_\_\_\_\_ Make sure the computer is plugged in. Make sure the power source is on.
- \_\_\_\_\_ Go back over the hardware installation to make sure that the system is properly installed.
- \_\_\_\_\_ Go back over the software installation to make sure you have installed all necessary software.
- \_\_\_\_\_ Run the Installation User Test to verify correct installation of both hardware and software.
- \_\_\_\_\_ Run the user-diagnostics test for your main board to make sure it's working properly.
- \_\_\_\_\_ Insert the FastVision CD-ROM and check the various Release Notes to see if there is any information relevant to the problem you are experiencing.

The release notes are available in the directory: `\usr\fastvision\alinfo`

## **20 FASTVISION TECHNICAL SUPPORT**

FastVision offers technical support to any licensed user during the normal business hours of 9 a.m. to 5 p.m. EST. We offer assistance on all aspects of processor board and PMC installation and operation.

## **21 CONTACTING TECHNICAL SUPPORT**

To speak with a Technical Support Representative on the telephone, call the number below and ask for Technical Support:

Telephone: **603-891-4317**

If you would rather FAX a written description of the problem, make sure you address the FAX to Technical Support and send it to:

Fax: **603-891-1881**

You can email a description of the problem to [support@fast-vision.com](mailto:support@fast-vision.com)

## **21.1 Returning Products for Repair or Replacements**

Our first concern is that you be pleased with your FastVision products.

If, after trying everything you can do yourself, and after contacting FastVision Technical Support, you feel your hardware or software is not functioning properly, you can return the product to FastVision for service or replacement. Service or replacement may be covered by your warranty, depending upon your warranty. The first step is to call FastVision and request a "Return Materials Authorization" (RMA) number. This is the number assigned both to your returning product and to all records of your communications with Technical Support. When an FastVision technician receives your returned hardware or software he will match its RMA number to the on-file information you have given us, so he can solve the problem you've cited.

When calling for an RMA number, please have the following information ready:

- \_\_\_\_\_ Serial numbers and descriptions of product(s) being shipped back
- \_\_\_\_\_ A listing including revision numbers for all software, libraries, applications, daughter cards, etc.
- \_\_\_\_\_ A clear and detailed description of the problem and when it occurs
- \_\_\_\_\_ Exact code that will cause the failure
- \_\_\_\_\_ A description of any environmental condition that can cause the problem

All of this information will be logged into the RMA report so it's there for the technician when your product arrives at FastVision. Put boards inside their anti-static protective bags. Then pack the product(s) securely in the original shipping materials, if possible, and ship to:

**FastVision, LLC.  
71 Spitbrook Rd Suite 200  
Nashua, NH 03060  
USA**

**Clearly mark the outside of your package:**

**Attention RMA 90XXX**

Remember to include your return address and the name and number of the person who should be contacted if we have questions.

## 21.2 Reporting Bugs

We at FastVision are continually improving our products to ensure the success of your projects. In addition to ongoing improvements, every FastVision product is put through extensive and varied testing. Even so, occasionally situations can come up in the fields that were not encountered during our testing at FastVision.

If you encounter a software or hardware problem or anomaly, please contact us immediately for assistance. If a fix is not available right away, often we can devise a work-around that allows you to move forward with your project while we continue to work on the problem you've encountered.

It is important that we are able to reproduce your error in an isolated test case. You can help if you create a stand-alone code module that is isolated from your application and yet clearly demonstrates the anomaly or flaw.

Describe the error that occurs with the particular code module and email the file to us at:

[support@fast-vision.com](mailto:support@fast-vision.com)

We will compile and run the module to track down the anomaly you've found.

If you do not have Internet access, or if it is inconvenient for you to get to access, copy the code to a disk, describe the error, and mail the disk to Technical Support at the FastVision address below.

If the code is small enough, you can also:

FAX the code module to us at 603-891-1881

If you are faxing the code, write everything large and legibly and remember to include your description of the error.

When you are describing a software problem, include revision numbers of all associated software.

For documentation errors, photocopy the passages in question, mark on the page the number and title of the manual, and either FAX or mail the photocopy to FastVision.

Remember to include the name and telephone number of the person we should contact if we have questions.

FastVision LLC.  
131 Daniel Webster Highway #529  
Nashua, NH 03060  
USA

Telephone: 603-891-4317  
FAX: 603-891-1881

Web site:  
<http://www.fast-vision.com/>

Electronic Mail:  
[sales@fast-vision.com](mailto:sales@fast-vision.com)  
[support@fast-vision.com](mailto:support@fast-vision.com)